

# Double-layered distributed transient frequency control with regional coordination

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**Abstract**—This paper proposes a control strategy for power systems with a two-layer structure that achieves global stabilization and, at the same time, delimits the transient frequencies of targeted buses to a desired safe interval. The first layer is a model predictive control that, in a receding horizon fashion, optimally allocates the power resources while softly respecting transient frequency constraints. As the first layer control requires solving an optimization problem online, it only periodically samples the system state and updates its action. The second layer control, however, is implemented in continuous time, assisting the first layer to achieve frequency invariance and attractivity requirements. Furthermore, through network partition, they can be implemented in a distributed fashion, only requiring system information from neighboring partitions. Simulations on the IEEE 39-bus network illustrate our results.

## I. INTRODUCTION

Power network frequency is used as a key performance metric in designing load shedding schemes [1]. In simulations, this frequency refers to a weighted average of the frequencies of all synchronous generators; however, in practice, due to the lack of availability of measurements for all generators, only a few of them are selected and sampled for monitoring and control design [2]. From the point of view of contingency recovery, even if the power supply and demand are re-balanced after a failure, due to the interconnected dynamics and inertia of power networks, individual buses may still be isolated from the network due to overheating relay protection. Therefore, there is a need of designing control schemes to restrict single bus transient frequency to evolve within an allowable range under disturbances and contingencies. This is the problem we address in this paper, paying attention to the distributed character of the controller and the reduction of the control effort through cooperation.

*Literature review:* The works [3], [4] propose sufficient conditions for power network synchronization; however, as bus transient frequency limits are not considered as constraints, the ideal synchronization condition may not hold due to possible violation in frequency transients. The work [5] studies the relation between power injection disturbance and frequency overshoot of individual bus without active control to regulate frequency transients. On the other hand, to actively control power network transients, several strategies have been investigated, including inertial placement [6], power system stabilizer [7], and power supply re-allocation [8]. Yet, these strategies, aiming at improving system transient behaviors, cannot rigorously constrain the evolution of frequency to stay within a safe region. In our

previous work [9], [10], we have developed two different control frameworks to achieve synchronization and frequency safety. The work [9] proposed a distributed controller, which is further enhanced in [10] to enable cooperation among neighboring buses in order to reduce the overall control effort in a receding horizon fashion. However, this control framework faces practical challenges for practical implementation because it takes a long period of time to find the optimal control trajectory and this must be done at every time instant. These limitations motivate our design here of a framework that can be implemented in real time while maintaining the advantage of cooperation.

*Statement of contribution:* This paper proposes a control strategy that achieves the following requirements through a dynamical state-feedback control design: (i) The closed-loop system is asymptotically stable. (ii) For every targeted bus, under perturbation from power injections or network dynamical interactions, its whole frequency trajectory stays within a given safe region, provided its initial frequency lies in the same region. (iii) If this is not the case, then the frequency trajectory should enter the safe region within a finite time and never leaves it afterwards. (iv) The control strategy is distributed by only requiring local state and network information. Hereby, we propose a double-layered control structure; by relaxing the frequency constraints and restricting the possible control trajectory from arbitrary to constant signal, the second-layer controller only needs to periodically (instead of continuously) solve an optimal control trajectory, and the time consumption for seeking the optimal one is greatly reduced and almost negligible. The first layer controller only slightly tunes the output of the second layer control signal so that the overall signal rigorously ensures requirement (i)-(iv). We also show that the proposed control is Lipschitz in state and continuous in time. For space reasons, all proofs are omitted and will appear elsewhere<sup>1</sup>.

<sup>1</sup>Throughout the paper, we employ the following notation. Let  $\mathbb{N}$ ,  $\mathbb{R}$ , and  $\mathbb{R}_{\geq}$  denote the set of natural, real, and nonnegative real numbers, resp. Variables belong to the Euclidean space unless specified otherwise.  $\mathbf{1}_n$  and  $\mathbf{0}_n$  in  $\mathbb{R}^n$  are the vector of all ones and zeros, resp. For  $a \in \mathbb{R}$ ,  $\lceil a \rceil$  denote its ceiling.  $\|\cdot\|$  denotes the 2-norm on  $\mathbb{R}^n$ . For  $b \in \mathbb{R}^n$ ,  $b_i$  denotes its  $i$ th entry. For  $A \in \mathbb{R}^{m \times n}$ , let  $[A]_i$  and  $[A]_{i,j}$  denote its  $i$ th row and  $(i,j)$ th element, resp. For any  $c, d \in \mathbb{N}$ , let  $[c, d]_{\mathbb{N}} = \{x \in \mathbb{N} | c \leq x \leq d\}$ . Denote the sign function  $\text{sgn} : \mathbb{R} \rightarrow \{0, 1\}$  as  $\text{sgn}(a) = 1$  if  $a \geq 0$ , and as  $\text{sgn}(a) = -1$  if  $a < 0$ . The saturation function  $\text{sat} : \mathbb{R} \rightarrow \mathbb{R}$  with limits  $a^{\min} < a^{\max}$  is  $\text{sat}(a; a^{\max}, a^{\min}) = a^{\max}$  if  $a \geq a^{\max}$ ,  $\text{sat}(a; a^{\max}, a^{\min}) = a^{\min}$  if  $a \leq a^{\min}$ , and  $\text{sat}(a; a^{\max}, a^{\min}) = a$  otherwise. An undirected graph [11] is a pair  $\mathcal{G} = (\mathcal{V}, \mathcal{E})$ , where  $\mathcal{V}$  is the vertex set and  $\mathcal{E} \subseteq \mathcal{V} \times \mathcal{V}$  is the edge set. An induced subgraph  $\mathcal{G}_{\beta} = (\mathcal{V}_{\beta}, \mathcal{E}_{\beta})$  of  $\mathcal{G}$  satisfies  $\mathcal{V}_{\beta} \subseteq \mathcal{V}$ ,  $\mathcal{E}_{\beta} \subseteq \mathcal{E}$ , and  $(i, j) \in \mathcal{E}_{\beta}$  if  $(i, j) \in \mathcal{E}$  with  $i, j \in \mathcal{V}_{\beta}$ . We denote by  $\mathcal{E}_{\beta}^i \subseteq \mathcal{V}_{\beta} \times (\mathcal{V} \setminus \mathcal{V}_{\beta})$  the edges connecting  $\mathcal{V}_{\beta}$  and the rest of the network. A graph is connected if there exists a path between any two vertices. For each edge  $e_k \in \mathcal{E}$  with vertices  $i, j$ , an orientation specifies either  $i$  or  $j$  as the positive end and the other as the negative end. The incidence matrix  $D = (d_{ki}) \in \mathbb{R}^{m \times n}$  associated with  $\mathcal{G}$  is defined as  $d_{ki} = 1$  if  $i$  is the positive end of  $e_k$ ,  $d_{ki} = -1$  if  $i$  is the negative end of  $e_k$ , and  $d_{ki} = 0$  otherwise.

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## II. PROBLEM STATEMENT

In this section we introduce the dynamics of the power network and the control requirements.

### A. Power network model

The power network is modeled by a connected undirected graph  $\mathcal{G} = (\mathcal{S}, \mathcal{E})$ , where  $\mathcal{S} = \{1, 2, \dots, n\}$  stands for the collection of buses (nodes) and  $\mathcal{E} = \{e_1, e_2, \dots, e_m\} \subseteq \mathcal{S} \times \mathcal{S}$  represents the collection of transmission lines (edges). For every bus  $i \in \mathcal{S}$ , let  $\omega_i \in \mathbb{R}$ ,  $p_i \in \mathbb{R}$ ,  $M_i \in \mathbb{R}_{\geq}$ , and  $E_i \in \mathbb{R}_{\geq}$  denote the nodal information of shifted voltage frequency relative to the nominal frequency, active power injection, inertial, and damping coefficient, resp. For simplicity, we assume that the latter two are strictly positive. Given an arbitrary orientation on  $\mathcal{G}$ , for any edge with positive end  $i$  and negative end  $j$ , let  $f_{ij}$  be its signed power flow and  $b_{ij} \in \mathbb{R}_{\geq}$  the line susceptance. Let  $\mathcal{S}^u \subset \mathcal{S}$  be the collection of buses with exogenous control inputs. To stack this notation in a more compact way, let  $f \in \mathbb{R}^m$ ,  $\omega \in \mathbb{R}^n$  and  $p \in \mathbb{R}^n$  denote the collection of  $f_{ij}$ 's,  $\omega_i$ 's, and  $p_i$ 's, resp. Let  $Y_b \in \mathbb{R}^{m \times m}$  be the diagonal matrix whose  $k$ th diagonal entry is the susceptance of the transmission line  $e_k$  connecting  $i$  and  $j$ , i.e.,  $[Y_b]_{k,k} = b_{ij}$ . Let  $M \triangleq \text{diag}(M_1, M_2, \dots, M_n) \in \mathbb{R}^{n \times n}$ ,  $E \triangleq \text{diag}(E_1, E_2, \dots, E_n) \in \mathbb{R}^{n \times n}$ , and  $D \in \mathbb{R}^{m \times n}$  be the incidence matrix. The linearized network dynamics is [12], [13],

$$\dot{f}(t) = Y_b D \omega(t), \quad (1a)$$

$$M \dot{\omega}(t) = -E \omega(t) - D^T f(t) + p(t) + \alpha(t), \quad (1b)$$

where  $\alpha(t) \in \mathbb{A} \triangleq \{y \in \mathbb{R}^n \mid y_w = 0 \text{ for } w \in \mathcal{S} \setminus \mathcal{S}^u\}$ . For convenience, we use  $x \triangleq (f, \omega) \in \mathbb{R}^{m+n}$ . We adopt the following assumption on the power injections.

*Assumption 2.1: (Finite-time convergence of active power injection).* For each  $i \in \mathcal{S}$ ,  $p_i$  is piece-wise continuous and becomes constant (denoted by  $p_i^*$ ) after a finite time, i.e., there exists  $0 \leq \bar{t} < \infty$  such that  $p_i(t) = p_i^*$  for every  $i \in \mathcal{S}$  and every  $t \geq \bar{t}$ . Furthermore, the constant power injections are balanced, i.e.,  $\sum_{i \in \mathcal{S}} p_i^* = 0$ .

Note that Assumption 2.1 generalizes the power injection profile from the commonly used time-invariant case (e.g. [14], [15]) to the finite-time convergent case. Also, as our controller design here lies in the scope of primary and secondary control, we assume that the power injection designed by the tertiary control through economic dispatch is balanced after a finite time. Under Assumption 2.1, one can show [9] that, for the open-loop system (i.e., (1) with  $\alpha \equiv \mathbf{0}_n$ ), the trajectories  $(f(t), \omega(t))$  globally converges to the unique equilibrium point  $(f_\infty, \mathbf{0}_n)$ , where  $f_\infty$  is uniquely determined by the power injection profile and network parameters.

### B. Control requirements

Given any non-empty subset  $\mathcal{S}^\omega$  of  $\mathcal{S}^u$ , the designed closed-loop system should meet the following requirements.

(i) *Frequency invariance:* For each  $i \in \mathcal{S}^\omega$ , let  $\underline{\omega}_i \in \mathbb{R}$  and  $\bar{\omega}_i \in \mathbb{R}$  be lower and upper safe frequency bounds, with  $\underline{\omega}_i < \bar{\omega}_i$ . The trajectory of  $\omega_i$  must stay inside  $[\underline{\omega}_i, \bar{\omega}_i]$ , provided that its initial frequency  $\omega_i(0)$  lies inside  $[\underline{\omega}_i, \bar{\omega}_i]$ .

This requirement guarantees that every targeted frequency always evolves inside the safe region.

(ii) *Frequency attractivity:* For each  $i \in \mathcal{S}^\omega$ , if  $\omega_i(0) \notin [\underline{\omega}_i, \bar{\omega}_i]$ , then there exists a finite time  $t_0$  such that  $\omega_i(t) \in [\underline{\omega}_i, \bar{\omega}_i]$  for every  $t \geq t_0$ . This requirement guarantees safe recovery from an undesired initial frequency.

(iii) *Asymptotic stability:* The controller should only regulate the system's transients, i.e., the closed-loop system should globally converge to the same equilibrium point  $(f_\infty, \mathbf{0}_n)$  of the open-loop system.

(iv) *Lipschitz continuity:* The controller must have Lipschitz in its state argument. This suffices to ensure the existence and uniqueness of solution for the closed-loop system and, furthermore, guarantees that the control action is robust to state measurement errors.

(v) *Economic cooperation:* The individual controllers  $\alpha_i$ ,  $i \in \mathcal{S}^u$ , should cooperate with each other to reduce the overall control effort measure by two norm.

(vi) *Distributed nature:* Every individual controller can only utilize the state and power injection information within a local region designed by operator. This reflects a practical requirement for implementation in larger-scale power networks, in which case centralized control strategies depending on global information may face critical challenge for real-time execution.

## III. CENTRALIZED DOUBLE-LAYERED CONTROLLER

In this section, we introduce a centralized controller that achieves the requirements (i)-(v) identified in Section II-B. Based on this design, we later propose a distributed version that also achieves requirement (vi).

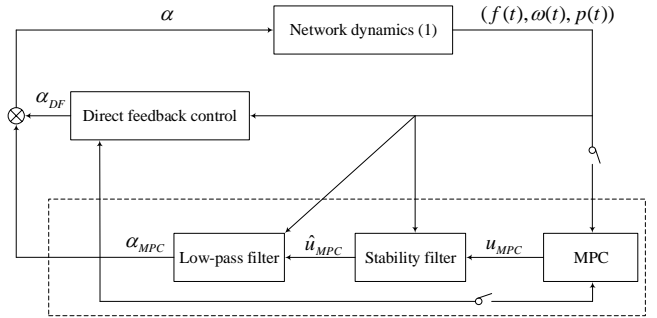


Fig. 1. Block diagram of the closed-loop system.

We adopt the centralized control structure depicted in Figure 1. The control signal  $\alpha$  consists of two parts

$$\alpha = \alpha_{DF} + \alpha_{MPC}. \quad (2)$$

We next describe the role played by each part. The bottom layer solves an optimization problem online. To do so, it combines an MPC component cascaded with a stability filter and a low-pass filter. The MPC component periodically and optimally allocates control resources, while roughly adjusting the frequency trajectories as a first step to achieve frequency invariance and attractivity. Its output is designed to be a

piece-wise constant signal  $u_{MPC}$ , which becomes a piece-wise continuous signal  $\hat{u}_{MPC}$  after passing through the stability filter. The low-pass filter ensures that the output  $\alpha_{MPC}$  of the bottom layer control is continuous in time to avoid any discontinuous change in control signal. Using real-time state information, the stability filter guarantees that  $\alpha_{MPC}$  does not jeopardize system stability. The bottom layer controller achieves economic cooperation and stabilization, but does not guarantee frequency invariance and attractivity. The top layer controller is called direct feedback control since, unlike the bottom layer control, can be directly computed in real time. This layer slightly modifies the control generated by the bottom layer to ensure frequency invariance and attractivity while maintaining stability of the system.

### A. Bottom layer controller design via MPC and filters

Here we formally describe each component in the bottom layer control and analyze their properties.

1) *MPC component*: The MPC component operates on a periodic time schedule. In each sampling period, the MPC component aims to allocate control resources over controlled nodes in an open-loop fashion based on the latest sampling system state and forecasted power injection. Here, due to the additional dynamics of the low-pass filter, the system state consists of not only power network state  $(f, \omega)$ , but also the state of low pass filter  $\alpha_{MPC}$  that we later explain. Formally, let  $\{\Delta^j\}_{j \in \mathbb{N}}$  be the collection of sampling points. At time  $t = \Delta^j$ , let a piece-wise continuous signal  $p_i^{fcst} : [t, t + \tilde{t}] \rightarrow \mathbb{R}^n$  be the forecasted value of the power injection  $p$  for the first  $\tilde{t}$  seconds after  $t$ . We discretize the dynamics (1) and denote  $N \triangleq \lceil \tilde{t}/T \rceil$  as the length of the predicted step with some  $T > 0$ . At each  $t = \Delta^j$ , the MPC component updates its output by solving the following optimization problem,

$$\begin{aligned} & \min_{\hat{f}, \hat{\omega}, \hat{\alpha}, \hat{u}, \beta} g(\hat{u}, \beta) \triangleq \sum_{i \in \mathcal{S}^u} c_i \hat{u}_i^2 + d\beta^2 \\ \text{s.t. } & \hat{f}(k+1) = \hat{f}(k) + TY_b D \hat{\omega}(k), \\ & M \hat{\omega}(k+1) = M \hat{\omega}(k) + T \{ -E \hat{\omega}(k) - D^T \hat{f}(k) + \\ & \quad \hat{p}^{fcst}(k) + \hat{u} \}, \forall k \in [0, N-1]_{\mathbb{N}}, \quad (3a) \\ & \hat{\alpha}_i(k+1) = \hat{\alpha}_i(k) + T \{ -\hat{\alpha}_i(k)/T_i - \hat{\omega}_i(k) + \hat{u}_i \}, \forall i \in \mathcal{S}^u, \\ & \hat{\alpha}_i \equiv 0, \quad \forall i \in \mathcal{S} \setminus \mathcal{S}^u, \quad (3b) \\ & \hat{u} \in \mathbb{A}, \quad (3c) \\ & \hat{f}(0) = f(\Delta^j), \quad \hat{\omega}(0) = \omega(\Delta^j), \quad \hat{\alpha}(0) = \alpha_{MPC}(\Delta^j), \quad (3d) \\ & \underline{\omega}_i - \beta \leq \hat{\omega}_i(k+1) \leq \bar{\omega}_i + \beta, \forall i \in \mathcal{S}^\omega, \forall k \in [0, N-1]_{\mathbb{N}}, \quad (3e) \\ & |\hat{u}_i| \leq \varepsilon_i |\alpha_{MPC, i}(\Delta^j)|, \quad \forall i \in \mathcal{S}^u. \quad (3f) \end{aligned}$$

In this optimization problem, (3a) is the discretized dynamics corresponding to (1) via first-order discretization, and  $\hat{p}^{fcst}(k) \triangleq p_{\Delta^j}^{fcst}(\Delta^j + kT)$  for every  $k \in [0, N-1]_{\mathbb{N}}$ ; (3b) is the discretized dynamics of the low-pass filter (explained below), with  $T_i > 0$  determining the filter bandwidth; (3c) indicates the availability of control signal indexes; (3d) is the initial state, where  $f(\Delta^j)$ ,  $\omega(\Delta^j)$ , and  $\alpha_{MPC}(\Delta^j)$  are sampled state values at time  $t = \Delta^j$ ; (3e) represents the relaxed constraint on frequency invariance, where we allow the discretized frequency  $\hat{\omega}_i$  with  $i \in \mathcal{S}^\omega$  exceed its bounds  $\underline{\omega}_i$  and  $\bar{\omega}_i$  at

the cost of a penalty term  $\beta$ ; (3f) bounds the control input  $\hat{u}_i$  via a coefficient  $\varepsilon_i > 0$  as a function of the state of the low-pass filter to limit the sensitivity to changes in the latter; the cost function  $g$  consists of the overall control effort as well as a penalty term for frequency violation, where  $c_i > 0$  for each  $i \in \mathcal{S}^u$  and  $d > 0$ . In the above expression, we use the compact notation  $\hat{F} \triangleq [\hat{f}(0), \hat{f}(1), \dots, \hat{f}(N)]$ ,  $\hat{\Omega} \triangleq [\hat{\omega}(0), \hat{\omega}(1), \dots, \hat{\omega}(N)]$ ,  $\hat{A} \triangleq [\hat{\alpha}(0), \hat{\alpha}(1), \dots, \hat{\alpha}(N)]$ ,  $\hat{P}^{fcst} \triangleq [\hat{p}^{fcst}(0), \hat{p}^{fcst}(1), \dots, \hat{p}^{fcst}(N-1)]$ .

We let  $\mathbf{R}(\mathcal{G}, \mathcal{S}^u, \mathcal{S}^\omega, p_{\Delta^j}^{fcst}, f(\Delta^j), \omega(\Delta^j), \alpha_{MPC}(\Delta^j))$  denote the optimization problem (3) to emphasize its dependence on network topology, nodal indexes with exogenous control signals, nodal indexes with transient frequency requirement, forecasted power injection, and state values at the sampling time. If the context is clear, we simply use  $\mathbf{R}$ . We let  $(\hat{F}^*, \hat{\Omega}^*, \hat{A}^*, \hat{u}^*, \beta^*)$  denote its optimal solution.

Given the open-loop optimization problem (3), the function  $u_{MPC}$  corresponding to the MPC component in Figure 1 is defined as follows: for  $j \in \mathbb{N}$  and  $t \in [\Delta^j, \Delta^{j+1})$ , let

$$u_{MPC}(t) = \hat{u}^*(\mathcal{G}, \mathcal{S}^u, \mathcal{S}^\omega, \hat{p}_{\Delta^j}^{fcst}, f(\Delta^j), \omega(\Delta^j), \alpha_{MPC}(\Delta^j)), \quad (4)$$

where in the right hand side we emphasize the dependence of  $\hat{u}^*$  on the seven arguments. Next, we characterize how the controller depends on the state value at the sampling time and predicted power injection.

*Lemma 3.1: (Piece-wise affine and continuous dependence of optimal solution on sampling state and predicted power injection).* The optimization problem  $\mathbf{R}(\mathcal{G}, \mathcal{S}^u, \mathcal{S}^\omega, p_{\Delta^j}^{fcst}, f(\Delta^j), \omega(\Delta^j), \alpha_{MPC}(\Delta^j))$  in (3) has a unique optimal solution  $(\hat{F}^*, \hat{\Omega}^*, \hat{A}^*, \hat{u}^*, \beta^*)$ . Furthermore, given  $\mathcal{G}$ ,  $\mathcal{S}^u$ , and  $\mathcal{S}^\omega$ ,  $\hat{u}^*$  is a continuous and piece-wise affine in  $(\hat{P}^{fcst}, f(\Delta^j), \omega(\Delta^j), \alpha_{MPC}(\Delta^j))$ , that is, there exist  $l \in \mathbb{N}$ ,  $\{H_i\}_{i=1}^l$ ,  $\{S_i\}_{i=1}^l$ ,  $\{h\}_{i=1}^l$ , and  $\{s_i\}_{i=1}^l$  with suitable dimensions such that

$$\hat{u}^* = S_i z + s_i, \text{ if } z \in \{y | H_i y \leq h_i\} \text{ and } i \in [1, l]_{\mathbb{N}} \quad (5)$$

holds for every  $z \in \mathbb{R}^{(N+2)n+m}$ , where  $z$  is the collection of  $(\hat{P}^{fcst}, f(\Delta^j), \omega(\Delta^j), \alpha_{MPC}(\Delta^j))$  in a column vector form.

Notice that Lemma 3.1 implies that  $\hat{u}^*$  is globally Lipschitz in  $z$  (and hence in the sampled state  $f(\Delta^j), \omega(\Delta^j)$ , and  $\alpha_{MPC}(\Delta^j)$ ), with  $L \triangleq \max_{i \in [1, l]_{\mathbb{N}}} \|F_i\|$  serving as a global Lipschitz constant. Another interesting consequence of this result is that it provides an alternative to directly solving the optimization problem  $\mathbf{R}$ . In fact, one can compute and store offline  $\{H_i\}_{i=1}^l$ ,  $\{S_i\}_{i=1}^l$ ,  $\{h\}_{i=1}^l$ , and  $\{s_i\}_{i=1}^l$ , and then compute  $\hat{u}^*$  online using (5). However, this approach faces practical difficulties regarding storage capacity [16], as the number  $l$  grows exponentially with system order  $m+n$ , input size  $|\mathcal{S}^u|$ , as well as the horizon length  $N$ .

2) *Stability and low-pass filter*: Next we introduce the stability and low-pass filters. Note that for any time  $t \in (\Delta_j, \Delta_{j+1})$ , due to the sampling mechanism,  $u_{MPC}(t)$  depends on the old sampled state at time  $\Delta_j$ , as opposed to the state information at current time  $t$ . Since such a lack of update may jeopardize system stability, we cascade a stability filter that depends on the current state after the MPC component

to filter out the unstable part in  $u_{MPC}$ . The goal of low-pass filter is to simply ensure that the output of the bottom layer is continuous in time. Formally, for every  $i \in \mathcal{S}^u$  at any  $t \geq 0$ , define the stability filter as

$$\begin{aligned} \hat{u}_{MPC,i}(\alpha_{MPC}(t), u_{MPC}(t)) \\ = \text{sat}(u_{MPC,i}(t); \varepsilon_i |\alpha_{MPC,i}(t)|, -\varepsilon_i |\alpha_{MPC,i}(t)|), \end{aligned} \quad (6)$$

and define the low-pass filter as

$$\begin{aligned} \dot{\alpha}_{MPC,i}(t) &= -\frac{1}{T_i} \alpha_{MPC,i}(t) - \omega_i(t) + \hat{u}_{MPC,i}(t), \quad \forall i \in \mathcal{S}^u, \\ \alpha_{MPC,i} &\equiv 0, \quad \forall i \in \mathcal{S} \setminus \mathcal{S}^u. \end{aligned} \quad (7)$$

Note that the low-pass filter model matches the structure in the discretized model (3b). Also, both (6) and (7) can be implemented in a distributed fashion:  $\alpha_{MPC,i}$  depends on  $\omega_i$  and  $\hat{u}_{MPC,i}$ , and  $\hat{u}_{MPC,i}$  only relies on  $\alpha_{MPC,i}$  and  $u_{MPC,i}$ , both of which are local information for node  $i$ . For simplicity, we interchangeably use  $\hat{u}_{MPC,i}(\alpha_{MPC}(t), u_{MPC}(t))$  and  $\hat{u}_{MPC,i}(t)$ .

The following result shows the Lipschitz continuity of  $\hat{u}_{MPC}$  and points out a condition it satisfies. This condition ensures asymptotic stability.

*Lemma 3.2: (Lipschitz continuity and stability condition).* For the signal  $\hat{u}_{MPC}$  defined in (6),  $\hat{u}_{MPC}$  is Lipschitz in system state at every sampling time  $t = \Delta^j$  with  $j \in \mathbb{N}$ . Additionally, it holds that  $u_{MPC,i}(t) \leq \varepsilon_i \alpha_{MPC,i}^2(t)$ ,  $\forall t \geq 0$ ,  $\forall i \in \mathcal{S}$ .

By Lemma 3.2, since  $\hat{u}_{MPC}$  is Lipschitz at every sampling time, if the top layer controller is also Lipschitz (which is the case, see Section III-B), then the solution of the closed-loop system exists and is unique and continuous in time. By (6) and noting that  $u_{MPC}$  is piece-wise constant in time, one has  $\hat{u}_{MPC}$  is piece-wise continuous, which further implies that  $\alpha_{MPC}$  is indeed continuous in time.

### B. Top layer design through direct feedback

Although the bottom layer control attempts to achieve frequency invariance and attractivity requirements by constraining the predicted frequency trajectory via (3e), it cannot solely guarantee the two requirements. To address this aspect, we construct the top layer control from [9] that is precisely designed to correct potential violations of the frequency requirements by kicking in as the margin of violations gets smaller. Formally, for every  $i \in \mathcal{S}^\omega$ , let  $\tilde{\gamma}_i, \underline{\gamma}_i > 0$ , and  $\underline{\omega}_i^{\text{thr}}, \bar{\omega}_i^{\text{thr}} \in \mathbb{R}$  with  $\underline{\omega}_i < \underline{\omega}_i^{\text{thr}} < 0 < \bar{\omega}_i^{\text{thr}} < \bar{\omega}_i$ . Define the top layer controller  $\alpha_{DF}$  as in (8).

Note that the top layer control signal is only available for node with index in  $\mathcal{S}^\omega$ , and that  $\alpha_{DF}$  can be implemented in a distributed way, in that for each  $\alpha_{DF,i}$  with  $i \in \mathcal{S}^\omega$  regulated at node  $i$ , it only requires its nodal frequency  $\omega_i$ , aggregated power flow  $[D^T]_i f$ , power injection  $p_i$ , as well as the local bottom layer control signal  $\alpha_{MPC,i}$ . In addition, we have shown [9] that  $\alpha_{DF}$  is locally Lipschitz in its first argument. If the context is clear, we may interchangeably use  $\alpha_{DF,i}(x(t), p(t), \alpha_{MPC}(t))$  (resp.  $v_i(x(t), \alpha_{MPC}(t), p(t))$ ) and  $\alpha_{DF,i}(t)$  (resp.  $v_i(t)$ ).

### C. Closed-loop stability, frequency invariance, and frequency attractivity analysis

With both layers introduced, we are ready to analyze the stability of the closed-loop system and show that it meets the requirements (i)-(iii) in Section II-B. Notice that as we individually introduce each component in the control scheme, we have shown that all components are Lipschitz, and the economic cooperation is encoded in the MPC component; therefore, the requirements (iv) and (v) are met.

*Theorem 3.3: (Centralized double-layered control with stability and frequency guarantees).* Under Assumption 2.1, if  $\varepsilon_i T_i < 1$  for every  $i \in \mathcal{S}^u$ , then the system (1) with controller defined by (2), (4), (6), (7), and (8) meets requirements (i)-(iii). Furthermore,  $\alpha(t)$ ,  $\alpha_{MPC}(t)$ , and  $\alpha_{DF}(t)$  converge to  $\mathbf{0}_n$  as  $t \rightarrow \infty$ .

Since the centralized double-layered control scheme already meets requirements (i)-(v), in the next section, we deal with the remaining distributed computation requirement.

## IV. CONTROLLER DECENTRALIZATION THROUGH NETWORK DIVISION

Going over the double-layered design in the previous section, it is worth noticing that the only component of the controller that requires global information is the MPC component, all the others being local in nature. In this section, we propose a distributed double-layered controller design that addresses this point. The general idea is to split the computation of the MPC component across different regions, and have each region determine its own MPC component based on its regional state and regional forecasted power information.

We split the network into regions so that each controlled node is contained in exactly one region. Formally, let  $\{\mathcal{G}_\beta = (\mathcal{S}_\beta, \mathcal{E}_\beta)\}_{\beta \in [1,d]_{\mathbb{N}}}$  be induced subgraphs of  $\mathcal{G}$  such that

$$\mathcal{S}^u \subseteq \bigcup_{\beta=1}^d \mathcal{S}_\beta, \quad (9a)$$

$$\mathcal{S}_\eta \cap \mathcal{S}_\beta \cap \mathcal{S}^u = \emptyset, \quad \forall \eta, \beta \in [1,d]_{\mathbb{N}} \text{ with } \eta \neq \beta. \quad (9b)$$

For each subgraph  $\mathcal{G}_\beta$ , let  $\mathcal{S}_\beta^u \triangleq \mathcal{S}^u \cap \mathcal{S}_\beta$  (resp.  $\mathcal{S}_\beta^\omega \triangleq \mathcal{S}^\omega \cap \mathcal{S}_\beta$ ) denote the collection of controlled node indexes (resp. nodes indexes with transient frequency requirements) within  $\mathcal{G}_\beta$ . Let  $(f_\beta, \omega_\beta, \alpha_{MPC,\beta}) \in \mathbb{R}^{2|\mathcal{S}_\beta| + |\mathcal{E}_\beta|}$  be the collection of states in  $\mathcal{G}_\beta$ . Let  $p_{t,\beta}^{\text{fctst}} : [t, t + \tilde{t}] \rightarrow \mathbb{R}^{|\mathcal{S}_\beta|}$  be the forecasted power injection for every node in  $\mathcal{G}_\beta$  starting from time  $t$  to  $\tilde{t}$  seconds later. Note that the dynamics of  $\mathcal{G}_\beta$  is not completely determined by  $(f_\beta, \omega_\beta, \alpha_{MPC,\beta})$  due to its interconnection with other parts of the network outside  $\mathcal{G}_\beta$  through transmission lines with  $i \in \mathcal{S}_\beta$  and  $j \in \mathcal{S} \setminus \mathcal{S}_\beta$  (equivalently, with  $(i, j) \in \mathcal{E}_\beta^t$ ). Instead of considering the flows  $f_{ij}$ 's of these transmission lines as states for  $\mathcal{G}_\beta$ , we model them as exogenous power injections. Formally, denote for every  $i \in \mathcal{S}_\beta$ ,

$$p_{t,\beta,i}^{\text{fctst},f}(\tau) \triangleq \sum_{\substack{j: j \rightarrow i \\ (j,i) \in \mathcal{E}_\beta^t}} f_{ji}(t) - \sum_{\substack{j: i \rightarrow j \\ (i,j) \in \mathcal{E}_\beta^t}} f_{ij}(t), \quad \forall \tau \in [t, t + \tilde{t}] \quad (10)$$

$$\forall i \in \mathcal{I}^\omega, \text{ let } \alpha_{DF,i}(x(t), p(t), \alpha_{MPC}(t)) = \begin{cases} \min\{0, \frac{\tilde{\gamma}_i(\bar{\omega}_i - \omega_i(t))}{\omega_i(t) - \bar{\omega}_i^{\text{thr}}} + v_i(x(t), \alpha_{MPC}(t), p(t))\} & \omega_i(t) > \bar{\omega}_i^{\text{thr}}, \\ 0 & \underline{\omega}_i^{\text{thr}} \leq \omega_i(t) \leq \bar{\omega}_i^{\text{thr}}, \\ \max\{0, \frac{\gamma_i(\omega_i - \underline{\omega}_i(t))}{\underline{\omega}_i^{\text{thr}} - \omega_i(t)} + v_i(x(t), \alpha_{MPC}(t), p(t))\} & \omega_i(t) < \underline{\omega}_i^{\text{thr}}, \end{cases} \quad (8a)$$

$$v_i(x(t), \alpha_{MPC}(t), p(t)) \triangleq E_i \omega_i(t) + [D^T]_i f(t) - p_i(t) - \alpha_{MPC,i}(t), \quad (8b)$$

$$\forall i \in \mathcal{I} \setminus \mathcal{I}^\omega, \text{ let } \alpha_{DF,i} \equiv 0. \quad (8c)$$

as the forecasted exogenous power injection acting on node  $i$  caused by transmission lines in  $\mathcal{E}_\beta^i$ , where  $\{j : j \rightarrow i\}$  is shorthand notation for  $\{j : j \text{ is the positive end of } (i, j)\}$ . For simplicity, here we take the forecasted value starting from time  $t$  to be constant within the time interval  $[t, t + \tilde{t}]$ . Denote by  $p_{t,\beta}^{fcst} : [t, t + \tilde{t}] \rightarrow \mathbb{R}^{|\mathcal{I}_\beta|}$  the collection of all  $p_{t,i}^{fcst}$ 's with  $i \in \mathcal{I}_\beta$ , and let  $\bar{p}_{t,\beta}^{fcst} \triangleq p_{t,\beta}^{fcst} + p_{t,\beta}^{fcst,f}$  be the overall forecasted power injection. We illustrate these definitions in an example.

*Example 4.1: (A network division in IEEE 39-bus network).* Fig. 2 shows a network division example with  $\mathcal{I}^\omega = \{30, 31, 32, 37\}$  and  $\mathcal{I}^u = \{3, 7, 25, 30, 31, 32, 37\}$ . The set  $\mathcal{I}^u$  consists of all nodes in  $\mathcal{I}^\omega$  and all nodes capable of adjusting their loads and within two hops of a node in  $\mathcal{I}^\omega$ . We split the network into three regions ( $d = 3$ ) satisfying (9). Each region  $\mathcal{G}_\beta$  contains the two-hop neighborhood for every node in  $\mathcal{I}_\beta^\omega$ . We denote by  $\mathcal{G}_1$  the upper left region in Fig. 2 and use it to illustrate related definitions. In  $\mathcal{G}_1$ , one has  $\mathcal{I}_1 = \{1, 2, 3, 25, 26, 30, 37\}$ ,  $\mathcal{I}_1^\omega = \{30, 37\}$ ,  $\mathcal{I}_1^u = \{3, 25, 30, 37\}$ ,  $\mathcal{E}_1 = \{(1, 2), (2, 30), (2, 25), (3, 25), (25, 37), (26, 37)\}$ , and  $\mathcal{E}_1^l = \{(1, 39), (3, 4), (3, 18), (26, 27), (26, 28), (26, 29)\}$ . For every  $i \in \mathcal{I}_1$ , one can compute  $p_{t,\beta,i}^{fcst,f}$  by (10), and it is easy to see that  $p_{t,\beta,i}^{fcst,f} \equiv 0$  for  $i \in \{2, 3, 25, 30, 37\}$ , as these nodes are not ends of any edge in  $\mathcal{E}_1^l$ . •

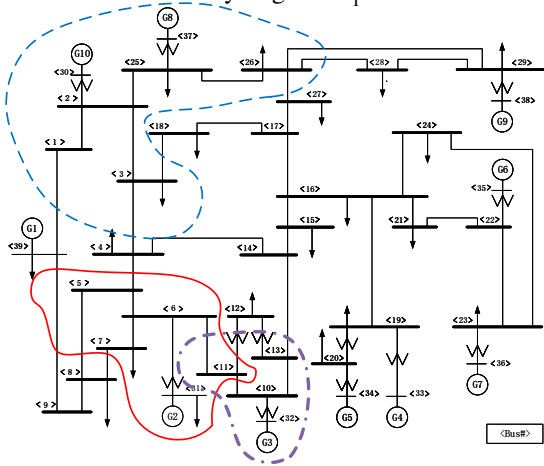


Fig. 2. IEEE 39-bus power network.

The key idea of designing the distributed MPC component is to consider each region as a single network and separately implement the centralized MPC on it. Formally, for every  $\beta \in [1, d]_{\mathbb{N}}$ , let  $\{\Delta_\beta^j\}_{j \in \mathbb{N}}$  be its sampling sequence. For every  $i \in \mathcal{I}^u$ , select the unique  $\beta$  such that  $i \in \mathcal{I}_\beta$ , and at every  $t \in [\Delta_\beta^j, \Delta_\beta^{j+1})$  with  $j \in \mathbb{N}$ , let

$$u_{MPC,i}(t)$$

$$= \hat{u}_i^*(\mathcal{G}_\beta, \mathcal{I}_\beta^u, \mathcal{I}_\beta^\omega, \bar{p}_{\Delta_\beta^j, \beta}^{fcst}, f_\beta(\Delta_\beta^j), \omega_\beta(\Delta_\beta^j), \alpha_\beta(\Delta_\beta^j)). \quad (11)$$

Compared to the centralized MPC in (4), the distributed version (11) transforms all global information, including network topology, forecasted power injection and system state, into local information. Their structural difference is that, in the distributed MPC, the overall forecasted power injection  $\bar{p}_{\Delta_\beta^j, \beta}^{fcst}$  includes an additional term (10) to account for the interconnected dynamics between the region of interest and the rest of the network. Next, we characterize the closed-loop stability and performance of the system under the distributed controller.

*Proposition 4.2: (Distributed double-layered control with stability and frequency guarantee).* Under Assumption 2.1 and assume that  $\varepsilon_i T_i < 1$  for every  $i \in \mathcal{I}^u$ , system (1) with controller defined by (2), (6), (7), (8), and (11) meets requirements (i)-(iii). Furthermore,  $\alpha(t)$ ,  $\alpha_{MPC}(t)$ , and  $\alpha_{DF}(t)$  converge to  $\mathbf{0}_n$  as  $t \rightarrow \infty$ .

## V. SIMULATIONS

We illustrate the performance of the distributed controller in the IEEE 39-bus power network described in Fig. 2. All parameters in the network model (1) are taken from the Power System Toolbox [17]. We assign a small rotational inertia  $M_i = 0.1$  to all non-generator nodes for simplicity. Let  $\bar{\omega}_i = -\underline{\omega}_i = 0.2 \text{ Hz}$ , so that the safe frequency region is  $[59.8 \text{ Hz}, 60.2 \text{ Hz}]$ . To set up the distributed MPC component (11), we select  $\tilde{t} = 2 \text{ s}$  and  $T = 0.02$ , so that the predicted step  $N = 100$ ;  $\varepsilon_i = 1.9$  and  $T_i = 0.5$  for every  $i \in \mathcal{I}^u$ ;  $c_i = 1$  if  $i \in \mathcal{I}^\omega$ , while  $c_i = 4$  if  $i \in \mathcal{I}^u \setminus \mathcal{I}^\omega$ ;  $d = 100$ ;  $\{\Delta_\beta^j\}_{j \in \mathbb{N}} = \{j\}_{j \in \mathbb{N}}$  for every  $\beta \in [1, d]_{\mathbb{N}}$ , i.e., in each region, the MPC component samples and updates its output every 1s;  $p_t^{fcst}(\tau) = p(\tau)$  for every  $\tau \in [t, t + \tilde{t}]$ . To set up the top layer controller (8), let  $\tilde{\gamma}_i = \gamma_i = 1$  and  $\bar{\omega}_i^{\text{thr}} = -\underline{\omega}_i^{\text{thr}} = 0.1 \text{ Hz}$  for every  $i \in \mathcal{I}^\omega$ .

We first show that the distributed controller defined by (2), (6), (7), (8), and (11) is able to maintain the targeted nodal frequencies within the safe region without changing the open loop equilibrium. We disturb all non-generator nodes by some time-varying power injections. In detail, for every  $i \in [1, 29]_{\mathbb{N}}$ , let  $p_i(t) = (1 + \delta(t))p_i(0)$ , where  $\delta(t) = 0.2 \sin(\pi t / 50)$  if  $0 \leq t \leq 25$ ;  $\delta(t) = 0.2$  if  $25 < t \leq 125$ ;  $\delta(t) = 0.2 \sin(\pi(t - 100) / 50)$  if  $125 < t \leq 150$ ;  $\delta(t) = 0$  if  $t > 150$ . The deviation term  $\delta(t)p_i(0)$  first drops down at a relatively fast rate and then remains steady for a long time period, finally converges to 0. We have chosen this scenario to test the capability of the controller against both fast and slow time-varying power injection disturbances.

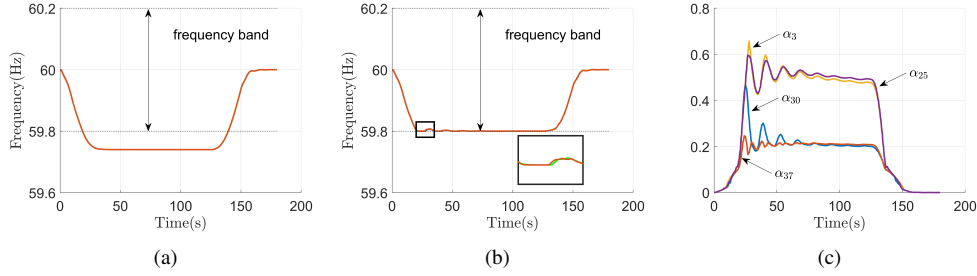


Fig. 3. Frequency and control input trajectories with and without distributed transient frequency control.

For simplicity, in the following we focus on the state and control input trajectories in the left-top region in Fig. 2. Fig. 3(a) shows the open-loop frequency responses of node 30 and 37, which have transient frequency requirements. The two nodes have almost the same overlapping trajectories that both exceed the safe lower frequency bound  $59.8\text{Hz}$ . As a comparison, in Fig. 3 (b), with the distributed controller, their frequency responses stay within the safe region, and also gradually come back to  $60\text{Hz}$  after the disturbance disappears. Given the selected coefficients  $c_3 = c_{25} = 1$  and  $c_{30} = c_{37} = 4$  in the optimization problem (3), the controller tends to use  $\alpha_3$  and  $\alpha_{25}$  more than  $\alpha_{30}$  and  $\alpha_{37}$ , and this is reflected in the control trajectories in Fig. 3(c).

Lastly, to verify that the proposed controller meets frequency attractivity requirement, we consider a case where the initial frequency is outside the safe region and see how the controller force the frequency back to the region. To do so, we disable in the setup above the distributed controller for the first 30s. In Fig. 4(a), one can see that the frequency of node 30 quickly recovers once we switch on the distributed controller. Fig. 4(b) shows the control signal of node 30. Note that, after transients,  $\alpha_{MPC,30}$  still dominates the overall control signal.

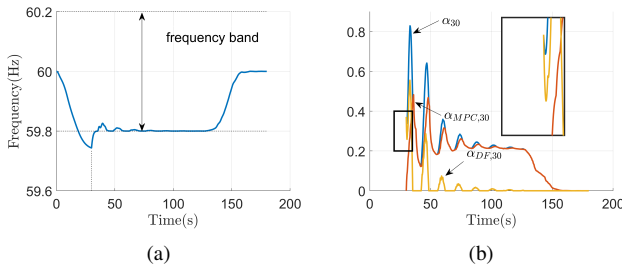


Fig. 4. Frequency and control input trajectories at node 30 with controller available after  $t = 30\text{s}$ .

## VI. CONCLUSIONS

We have proposed a distributed transient frequency control framework for power networks that preserves the asymptotic stability of the network and at the same time, guarantees safe frequency interval invariance and attractivity for targeted nodes. The controller possesses a double-layered structure, with the bottom layer periodically sampling the state and allocating control signals over a local region in a receding horizon fashion. The top layer slightly tunes the bottom layer signal in order to provably enforce frequency invariance

and attractivity guarantees. Implemented over a network partition, both layers rely on local state and power injection information. Future work will investigate the extension of the results to nonlinear swing dynamics, the optimal selection of sampling sequences in the bottom layer control design, the analysis of the performance trade-offs of the parameter selections, and the designs of distributed control schemes that do not rely on network partitions.

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