

MAE40 - Linear Circuits - Fall 20
Final Exam, December 19, 2020

Instructions

- (i) Prior to the exam, you must have completed the Academic Integrity Pledge at <https://academicintegrity.ucsd.edu/forms/form-pledge.html>
- (ii) The exam is open book. You may use your class notes and textbook
- (iii) Collaboration is not permitted. The answers you provide should be the result only of your own work
- (iv) On the questions for which the answers are given, please provide detailed derivations
- (v) The exam has 4 questions for a total of 40 points and 2 bonus points
- (vi) You have from 8:00am to 11:00am to complete the exam. Allow sufficient time to post your answers in Canvas (submission closes at 11:15am).
- (vii) If there is any clarification needed, post your question in the “Discussions” tab of the class Canvas webpage (“Clarifications on question statements of final”)

Good luck!

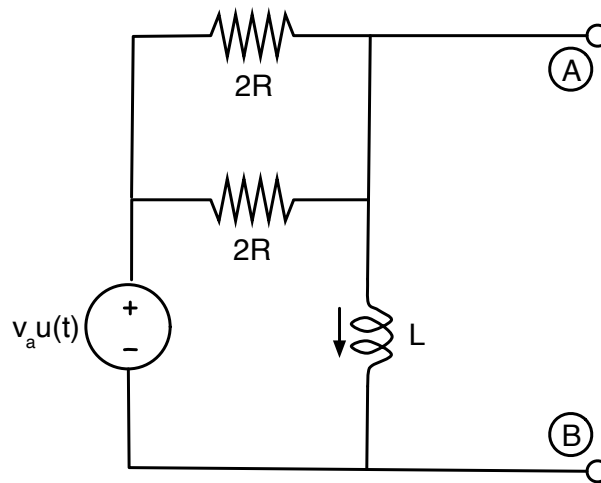


Figure 1: Circuit for Question 1.

1. Equivalent Circuits

Here, v_a is a constant.

Part I: [2 points] Assuming $i_L(0) = 1A$, transform the circuit in Figure 1 into the s -domain, using a voltage source to account for the initial condition of the inductor.

Part II: [2 points] For the circuit you obtained in Part I, find the open-circuit voltage transform as seen from terminals \textcircled{A} - \textcircled{B} . The answer should be given as a ratio of two polynomials.

Part III: [2 points] For the circuit you obtained in Part I, find the short-circuit current transform as seen from terminals \textcircled{A} - \textcircled{B} . The answer should be given as a ratio of two polynomials.

Part IV: [2 points] For the circuit you obtained in Part I, find the Thévenin equivalent in the s -domain as seen from terminals \textcircled{A} - \textcircled{B} (the impedance should be given as a ratio of two polynomials).

Part V: [2 points] Break down the open-circuit voltage transform you obtained in Part II as the sum of the zero-state and zero-input response transforms. Do the same as the sum of the forced and natural response transforms.

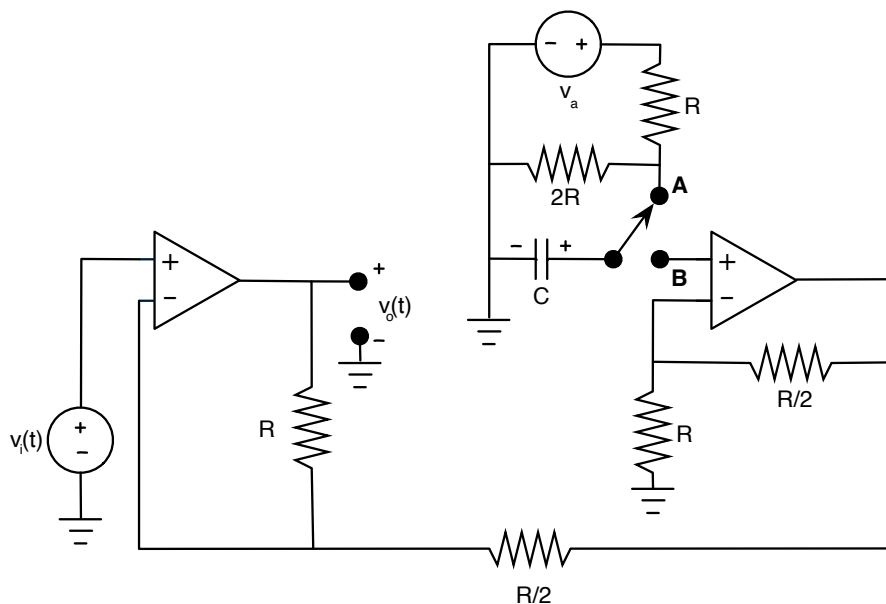


Figure 2: RCL circuit for Laplace Analysis for Question 2.

2. Laplace Domain Circuit Analysis

Part I: [2 points] Consider the circuit depicted in Figure 2. The value v_a of the current source is constant. The switch is kept in position **A** for a very long time. At $t = 0$, it is moved to position **B**. Show that the initial condition for the capacitor is given by

$$v_C(0^-) = \frac{2}{3}v_a.$$

[Show your work]

Part II: [5 points] Use this initial condition to transform the circuit into the s -domain for $t \geq 0$. Use an equivalent model for the capacitor in which the initial condition appears as a voltage source. Use nodal analysis to express the output response transform $V_o(s)$ as a function of $V_i(s)$ and v_a .

Part III: [3 points] Use partial fractions and inverse Laplace transforms to show that the output voltage $v_o(t)$ when $v_a = 2\text{ V}$, $v_i(t) = te^{-2t}u(t)\text{ V}$, $C = 10\text{ mF}$, and $R = 100\text{ K}\Omega$ is

$$v_o(t) = (3te^{-2t} - 4)u(t).$$

Part IV: [Extra 2 points] Decompose the output voltage of Part III as (i) the sum of the natural and forced response, and (ii) the sum of the zero-state and zero-input response.

3. Frequency Response Analysis

Consider the transfer function

$$T(s) = \frac{500s}{s^2 + 520s + 10^4}$$

Part I [3 points] Compute the gain $|T(j\omega)|$ and phase $\angle T(j\omega)$ functions

Part II [3 points] What are the DC gain and the ∞ -freq gain? What are the corresponding values of the phase function? What are the cut-off frequencies?

Part III [2 points] Sketch plots for the gain and phase functions. What type of filter is this one?
[Explain your answer]

Part IV [2 points] Using what you know about frequency response, compute the steady-state response $v_o^{SS}(t)$ to the input $v_i(t) = \cos(300t + \frac{\pi}{4})$.

4. OpAmp Design

Consider the transfer function

$$T(s) = \frac{500s}{s^2 + 520s + 10^4}$$

of the previous question.

Part I: [4 points] Consider the following factorization of the transfer function

$$T(s) = \frac{-20}{s + 20} \cdot \frac{-25s}{s + 500}$$

Based on this decomposition, design a circuit as the series connection of two stages, each with 1 OpAmp, that implements $T(s)$. Additionally, you can only use 100Ω and $1K\Omega$ -resistors, and $20\mu F$ and $50\mu F$ -capacitors, and no inductors. Be sure to properly justify why the overall transfer function of your design is the product of the individual transfer function of each stage.

Part II: [3 points] Consider instead the following factorization of the transfer function

$$T(s) = \frac{s}{s + 20} \cdot \frac{500}{s + 500}$$

Based on this decomposition, design a circuit as the series connection of three stages, one with 1 OpAmp and the others without, that implements $T(s)$. Additionally, you can only use 100Ω and $1K\Omega$ -resistors, and $20\mu F$ and $50\mu F$ -capacitors, and no inductors. Be sure to properly justify why the overall transfer function of your design is the product of the individual transfer function of each stage.

Part III: [3 points] Finally, consider the following factorization of the transfer function

$$T(s) = \frac{20}{s + 20} \cdot 25 \cdot \frac{s}{s + 500}$$

Based on this decomposition, design a circuit as the series connection of three stages, one with 1 OpAmp and the others without, that implements $T(s)$. Additionally, you can only use 100Ω and $1K\Omega$ -resistors, and $20\mu F$ and $50\mu F$ -capacitors, and no inductors. Be sure to properly justify why the overall transfer function of your design is the product of the individual transfer function of each stage.