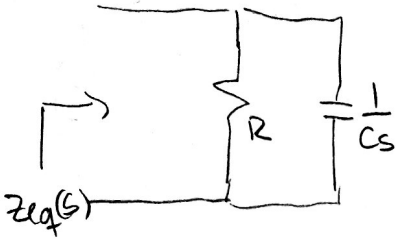


Homework 7

10.2)



$$Z_{eq} = \frac{(R) \left(\frac{1}{Cs} \right)}{R + \frac{1}{Cs}} = \frac{R}{RCs + 1}$$

For the pole $s = -250 \text{ krad/s}$ we have:

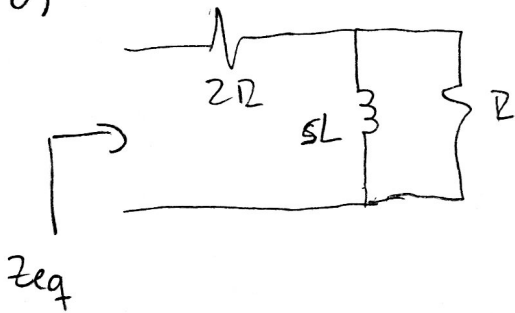
$$RCs + 1 = 0 \Rightarrow s = -\frac{1}{RC}, \text{ so we pick } C = 100 \text{ pF.}$$

It follows that

$$R = \frac{-1}{sC} = \frac{-1}{(-250 \text{ k}, 100 \text{ p})} = \frac{1}{(250 \times 10^3)(100 \times 10^{-12})}$$

$$= 40 \text{ k}\Omega$$

10.8)



a)

$$Z_{eq} = 2R + (sL \parallel R)$$

$$= 2R + \frac{sLR}{sL + R}$$

$$= \frac{2R(sL + R) + sLR}{sL + R}$$

$$= \frac{3sLR + 2R^2}{sL + R}$$

To find the poles, we have

$$sL + R = 0 \Rightarrow s = -\frac{R}{L}$$

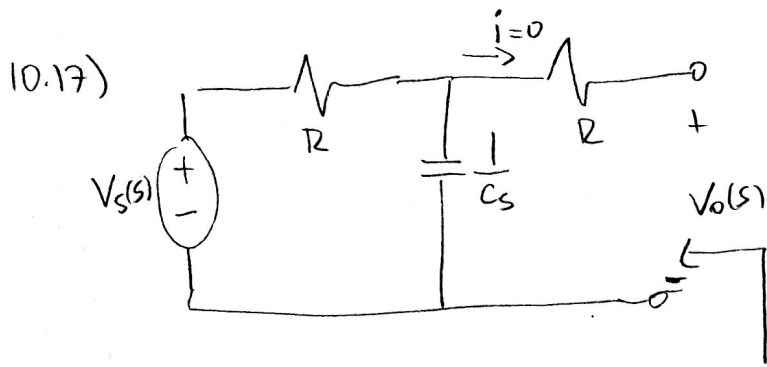
For zeros, $3sLR + 2R^2 = 0 \Rightarrow s = \frac{-2R^2}{3LR} = -\frac{2R}{3L}$

b) It is given a pole at -330 rad/s , it implies that $-330 = -\frac{R}{L}$.

Since the zero is located at $s = -\frac{2}{3} \frac{R}{L}$, we have that

$$s = -\frac{2}{3}(330) = -220, \text{ so the zero is located at } s = -220 \text{ rad/s.}$$

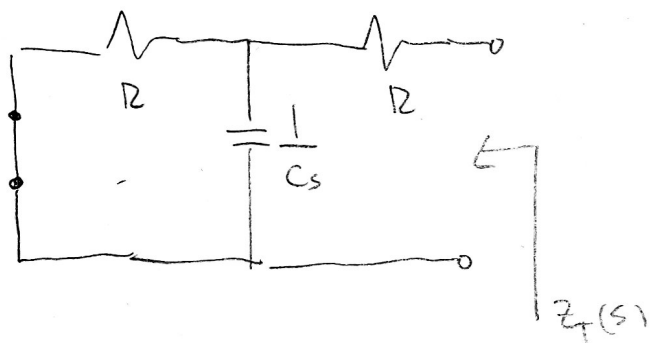
We pick $L = 10 \text{ H}$, so we have $R = (330)10 = 3.3 \text{ k}\Omega$.



a) Since $i=0$, $V_o(s)$ is the voltage in the capacitor, it follows

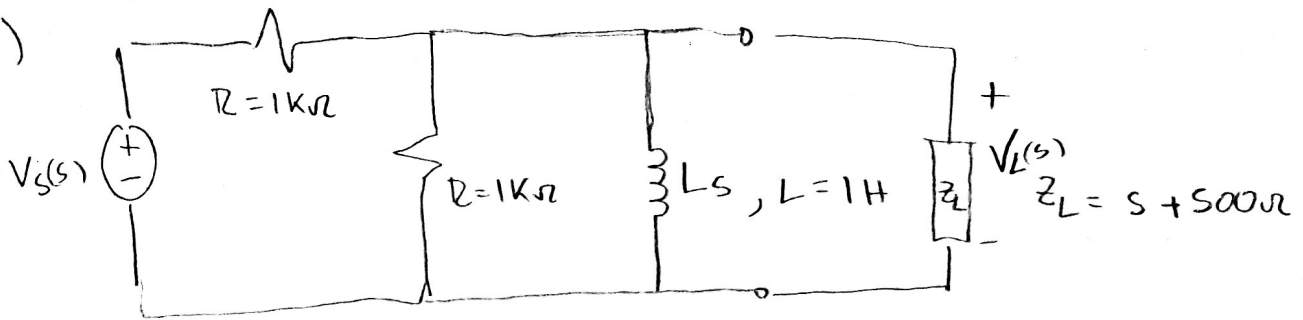
$$V_o(s) = \frac{\left(\frac{1}{c_s}\right)}{\frac{1}{c_s} + R} (V_s(s)) = \frac{V_s(s)}{1 + RCs}$$

b) We turn off all independent sources, i.e.,

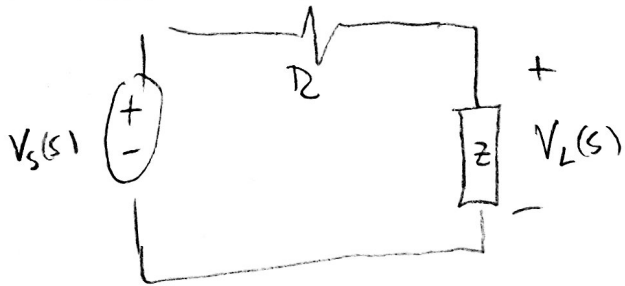


$$\begin{aligned} Z_T(s) &= R + \left(R \parallel \frac{1}{c_s}\right) \\ &= R + \frac{(R)\left(\frac{1}{c_s}\right)}{R + \frac{1}{c_s}} \\ &= R + \frac{R}{RCs + 1} \\ &= \frac{R^2 c_s + R + R}{RCs + 1} \\ &= \frac{R^2 c_s + 2R}{RCs + 1} \end{aligned}$$

10.22)



The circuit is equivalent to:



$$\begin{aligned} \text{where } Z &= R \parallel Ls \parallel Z_L = (R \parallel Ls \parallel (s+500)) = \left(\frac{1}{R} + \frac{1}{Ls} + \frac{1}{s+500} \right)^{-1} \\ &= \left(\frac{Ls(s+500) + R(s+500) + RLs}{RLs(s+500)} \right)^{-1} \\ &= \frac{RLs(s+500)}{Ls^2 + s(500L + R + RL) + 500R} \end{aligned}$$

Now, we use voltage division:

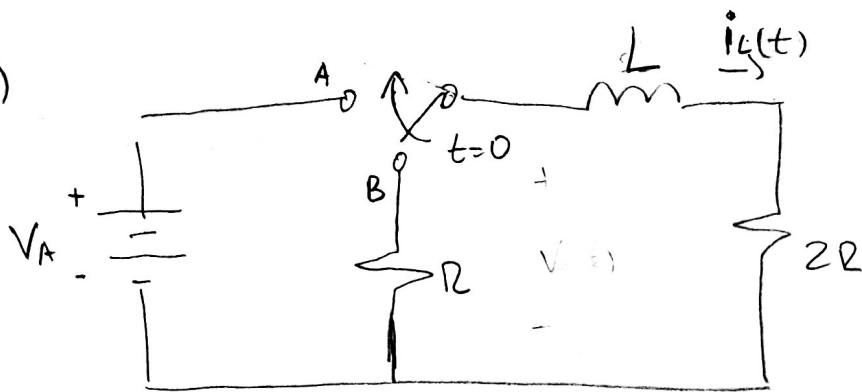
$$V_L(s) = \frac{Z}{Z+R} V_s(s) = \frac{(RLs^2 + 500RLs) V_s(s)}{(Ls^2 + (500L + R + RL)s + 500R) \left(\frac{RLs^2 + 500RLs}{Ls^2 + s(500L + R + RL) + 500R} + R \right)}$$

$$\begin{aligned} \text{(Replacing values and reorganizing)} &= \frac{(1k s^2 + 500k s) V_s(s)}{1k s^2 + 500k s + 1k(s^2 + (500 + 1k + 1k)s + 500k)} \\ &= \frac{(1k s^2 + 500k s) V_s(s)}{2k s^2 + (500k + 2500k)s + 500.000k} \\ &= \frac{\frac{1}{2} s (s + 500) V_s(s)}{s^2 + 1500s + 250 \cdot 10^3} \\ &= \frac{\frac{1}{2} s (s + 500) V_s(s)}{(s + 191)(s + 1309)} \end{aligned}$$

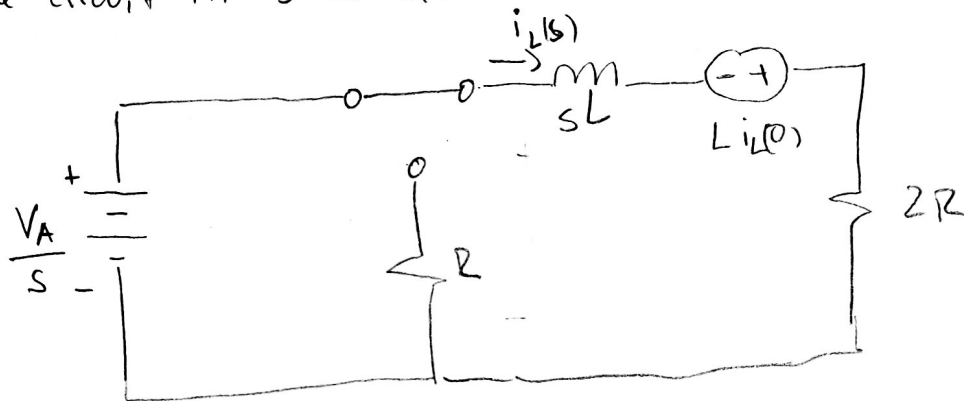
Natural poles: $s = -191 \text{ rad/s}$ and $s = -1309 \text{ rad/s}$.

Natural zeros: $s = 0$ and $s = -500 \text{ rad/s}$

10.26)

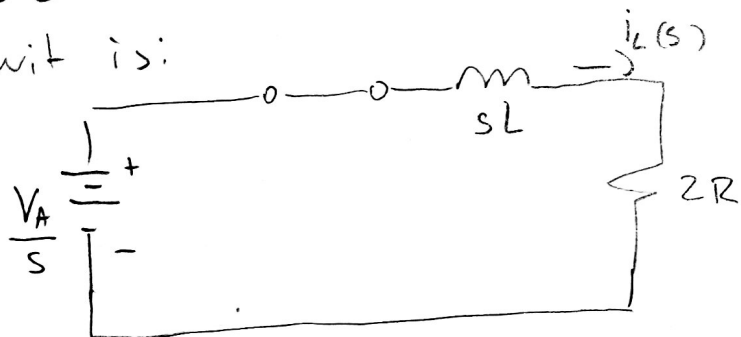


The circuit in s domain:



Since the position of the switch was in B for a long time before moved to position A, then $i_L(0) = 0$. The equivalent

circuit is:



$$i_L(s) = \frac{\frac{V_A}{s}}{sL + 2R} = \frac{V_A}{s(sL + 2R)} = \frac{\frac{V_A}{L}}{s(s + \frac{2R}{L})} = \frac{K_1}{s} + \frac{K_2}{s + \frac{2R}{L}}$$

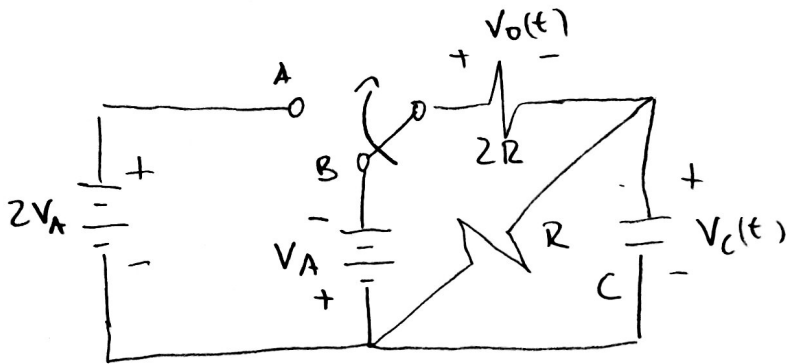
$$K_1 = s i_L(s) \Big|_{s=0} = \frac{\frac{V_A}{L}}{s + \frac{2R}{L}} \Big|_{s=0} = \frac{V_A}{2R}$$

$$K_2 = \left(s + \frac{2R}{L} \right) i_L(s) \Big|_{s = -\frac{2R}{L}} = \frac{\frac{V_A}{L}}{s} \Big|_{s = -\frac{2R}{L}} = -\frac{V_A}{2R}$$

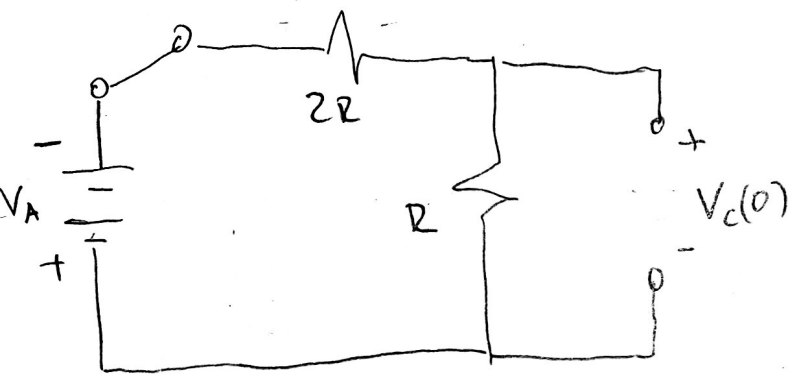
$$i_L(s) = \frac{\frac{V_A}{2R}}{s} - \frac{\frac{L V_A}{2R}}{s + \frac{2R}{L}}$$

$$i_L(t) = \frac{V_A}{2R} u(t) - \frac{V_A}{2R} e^{-\frac{2R}{L}t} u(t)$$

10.28)



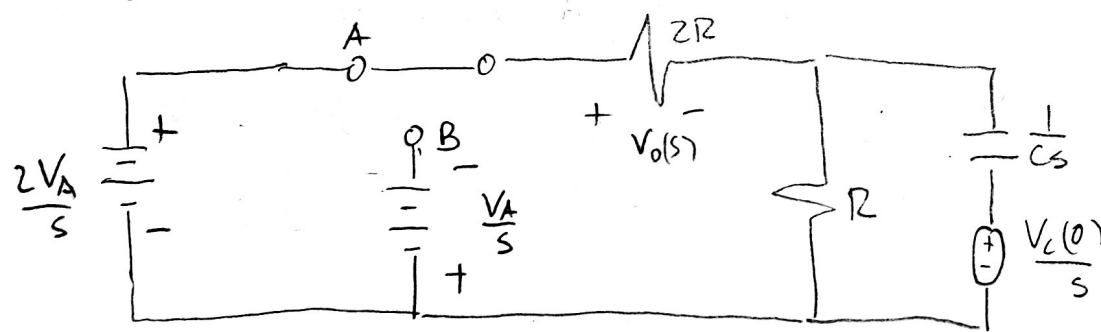
First, we find the initial condition when the switch is in position B for $t < 0$. The corresponding circuit is shown below.



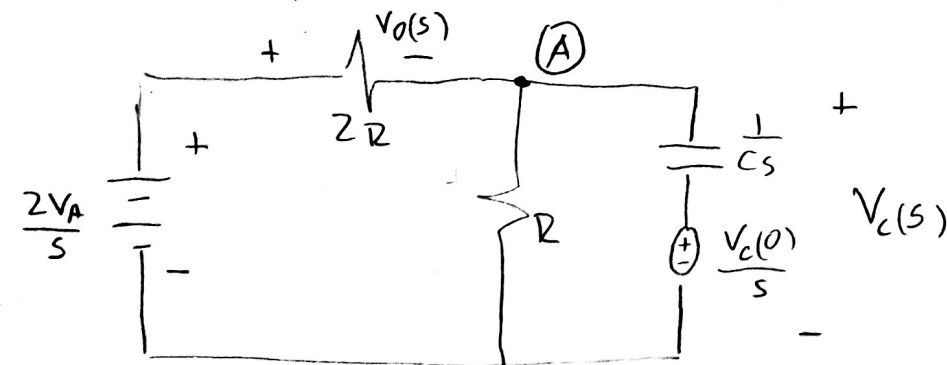
Since the circuit for $t < 0$ is in dc for long time, then the capacitor acts as an open circuit. It follows

$$V_c(0) = \frac{R}{3R} (-V_A) = -\frac{1}{3} V_A$$

Next, we show the circuit in s domain when $t \geq 0$



which is equivalent to:



Using KCL at node A, we have

$$\frac{\frac{2V_A}{s} - V_C(s)}{2R} - \frac{V_C(s)}{R} + \frac{V_C(s) + \frac{V_A}{3s}}{1/Cs} = 0$$

$$\left(\frac{1}{2R} + \frac{1}{R} + Cs\right) V_C(s) = \frac{2V_A}{2Rs} - \frac{C}{3} V_A$$

$$\left(\frac{3 + 2RCs}{2R}\right) V_C(s) = \left(\frac{3 - RCs}{3Rs}\right) V_A$$

$$V_C(s) = \frac{(6 - 2RCs)V_A}{9s + 6RCs^2} = \frac{-\frac{1}{3}\left(s - \frac{3}{RC}\right)V_A}{s\left(s + \frac{3}{2RC}\right)}$$

$$= \frac{K_1}{s} + \frac{K_2}{s + \frac{3}{2RC}}$$

$$K_1 = sV_C(s) \Big|_{s=0} = \frac{-\frac{1}{3}\left(-\frac{3}{RC}\right)V_A}{\frac{3}{2RC}} = \frac{2}{3} V_A$$

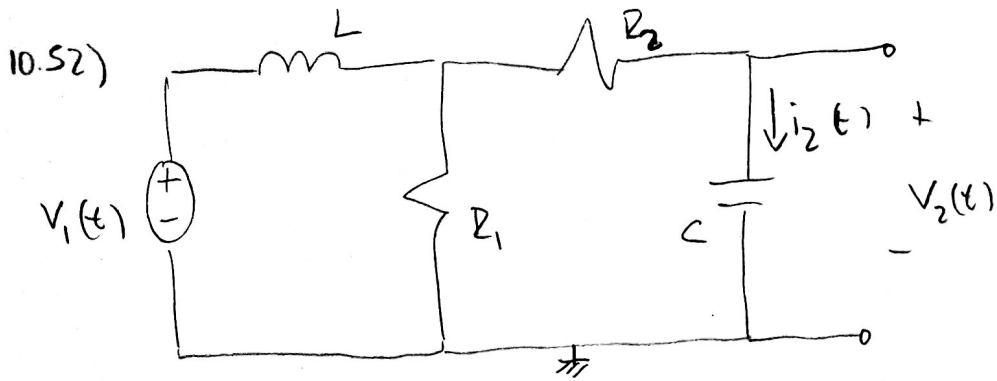
$$K_2 = \left(s + \frac{3}{2RC}\right) V_C(s) \Big|_{s = -\frac{3}{2RC}} = \frac{-\frac{1}{3}\left(-\frac{3}{2RC} - \frac{3}{2RC}\right)V_A}{-\frac{3}{2RC}} = \frac{-\left(\frac{1}{2} + 1\right)V_A}{\frac{3}{2}} = -V_A$$

$$V_C(s) = \frac{\frac{2}{3} V_A}{s} - \frac{V_A}{s + \frac{3}{2RC}}$$

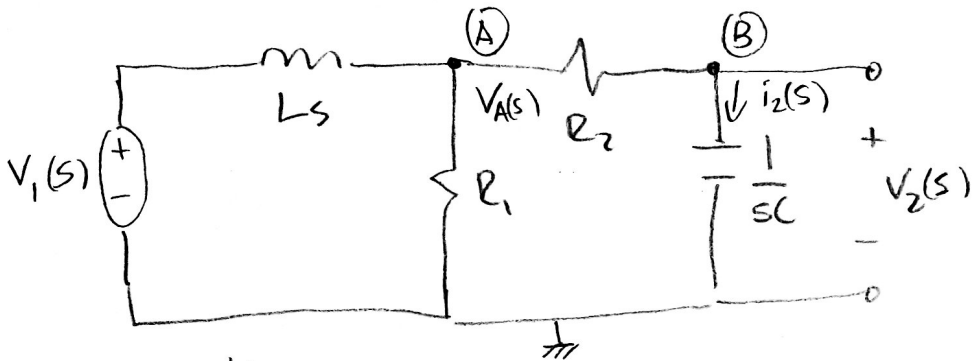
$$V_C(t) = \frac{2}{3} V_A u(t) - V_A e^{-\frac{3t}{2RC}} u(t)$$

$$V_O(s) = \frac{2V_A}{s} - V_C(s) = \frac{2V_A}{s} - \frac{\frac{2}{3} V_A}{s} + \frac{V_A}{s + \frac{3}{2RC}} = \frac{\frac{4}{3} V_A}{s} + \frac{V_A}{s + \frac{3}{2RC}}$$

$$V_O(t) = \frac{4}{3} V_A u(t) + V_A e^{-\frac{3t}{2RC}} u(t)$$



a) The circuit in s-domain with initial condition equal to zero?



Node-Voltage equations:

$$\left(\frac{1}{Ls} + \frac{1}{R_1} + \frac{1}{R_2} \right) V_A(s) - \frac{1}{R_2} V_2(s) = \frac{V_1(s)}{Ls} \quad (\text{Node A}) \quad (*)$$

$$-\frac{1}{R_2} V_A(s) + \left(\frac{1}{R_2} + sC \right) V_2(s) = 0 \quad (\text{Node B})$$

b)

$$V_A(s) = \left(\frac{1 + R_2 C s}{R_2} \right) R_2 V_2(s) \quad (\text{From node-voltage at node B})$$

we replace V_A in (*):

$$\left(\frac{R_1 R_2 + R_2 L s + R_1 L s}{R_1 R_2 L s} \right) (1 + R_2 C s) V_2(s) - \frac{1}{R_2} V_2(s) = \frac{V_1(s)}{Ls}$$

$$\left((R_1 R_2 + R_2 L s + R_1 L s)(1 + R_2 C s) - R_1 L s \right) V_2(s) = R_1 R_2 V_1(s)$$

$$V_2(s) = \frac{R_1 R_2 V_1(s)}{R_2 (R_1 + R_2) L C s^2 + (R_1 + R_2 - R_1) L s + R_1 R_2 C s + R_1 R_2}$$

$$= \frac{R_1 V_1(s)}{(R_1 + R_2) L (C s^2 + (R_1 R_2 C + L) s + R_1)}$$

10.52) cont:

c) The natural poles are located at

$$s = \frac{-(R_1 R_2 C + L) \pm \sqrt{(R_1 R_2 C + L)^2 - 4(R_1 + R_2)LC R_1}}{2(R_1 + R_2)LC}$$

d) $V_1(t) = 5u(t) \Rightarrow V_1(s) = \frac{5}{s}$, Replacing, we have

$$\begin{aligned} V_2(s) &= \frac{5R_1}{((R_1 + R_2)LCs^2 + (R_1 R_2 C + L)s + R_1)s} \\ &= \frac{5(500)}{s((1k)(0.5)(1\mu)s^2 + (500^2(1\mu) + 0.5)s + 500)} \\ &= \frac{2500}{s(500 \times 10^{-6}s^2 + 0.75s + 500)} \\ &= \frac{5 \times 10^6}{s(s^2 + 1500s + 10 \times 10^5)} = \frac{A}{s} + \frac{Bs + C}{s^2 + 1500s + 10 \times 10^5} \end{aligned}$$

$$5 \times 10^6 = A(s^2 + 1500s + 10 \times 10^5) + Bs^2 + Cs$$

$$A + B = 0 \Rightarrow B = -5$$

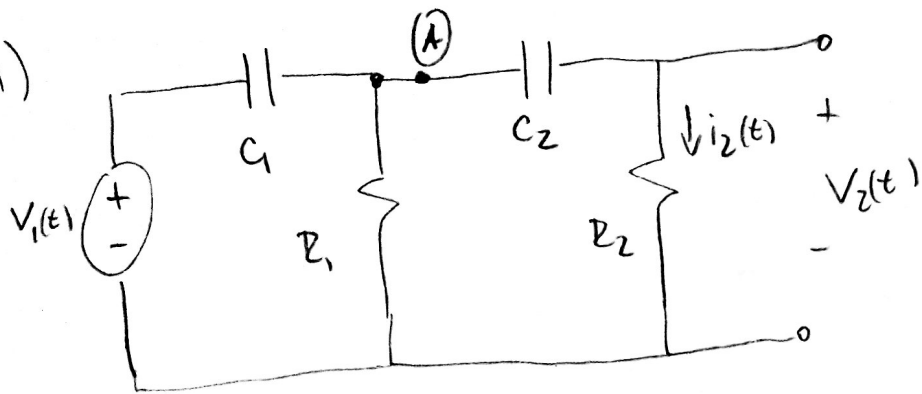
$$1500A + C = 0 \Rightarrow C = -7500$$

$$5 \times 10^6 = 10 \times 10^5 A \Rightarrow A = 5$$

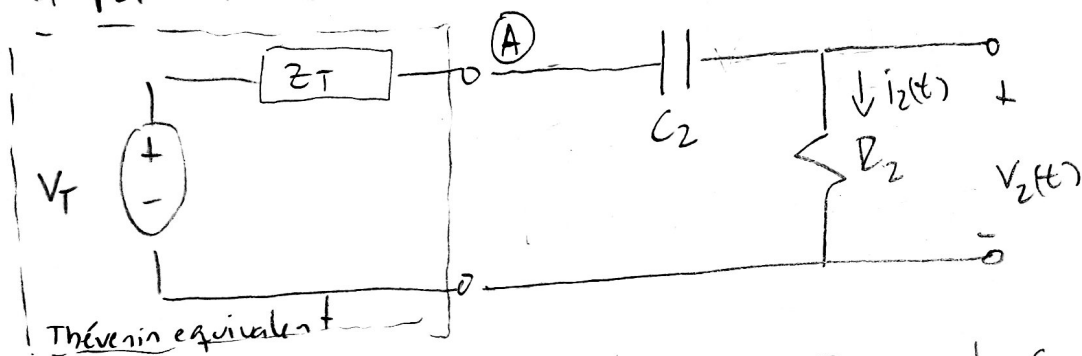
$$V_2(s) = \frac{5}{s} - \frac{5s + 7500}{s^2 + 1500s + 1 \times 10^6} = \frac{5}{s} - \frac{5(s + 1500)}{(s + 750)^2 + (661.44)^2}$$

$$V_2(t) = 5u(t) - 5e^{-750t} \cos(661.44t)u(t) - 5.67 e^{-750t} \sin(661.44t)u(t)$$

10.54)



When $V_1(t) = \frac{1}{s}$, the Thévenin equivalent circuit to the left of point A is shown below, with $Z_T = \frac{10^6}{s+10^3} \Omega$ and $V_T(s) = \frac{1}{s+10^3} V-s$.



We have to select the values of R_2 and C_2 s.t.:

$$V_2(s) = \frac{s}{s^2 + 3000s + 10^6}$$

We apply voltage division:

$$V_2(s) = \frac{R_2 V_T(s)}{Z_T + \frac{1}{C_2 s} + R_2} = \frac{R_2 \frac{1}{s+10^3}}{\frac{10^6}{s+10^3} + \frac{1}{C_2 s} + R_2}$$

$$= \frac{\frac{(s+10^3) R_2 s}{C_2 (s+10^3)}}{10^6 C_2 s + s+10^3 + R_2 C_2 (s+10^3) s} = \frac{s}{s^2 + (10^3 + \frac{10^6}{R_2} + \frac{1}{R_2 C_2}) s + \frac{10^3}{R_2 C_2}}$$

We need: $\left(10^3 + \frac{10^6}{R_2} + \frac{1}{R_2 C_2}\right) = 3000$ and $\frac{10^3}{R_2 C_2} = 10^6$

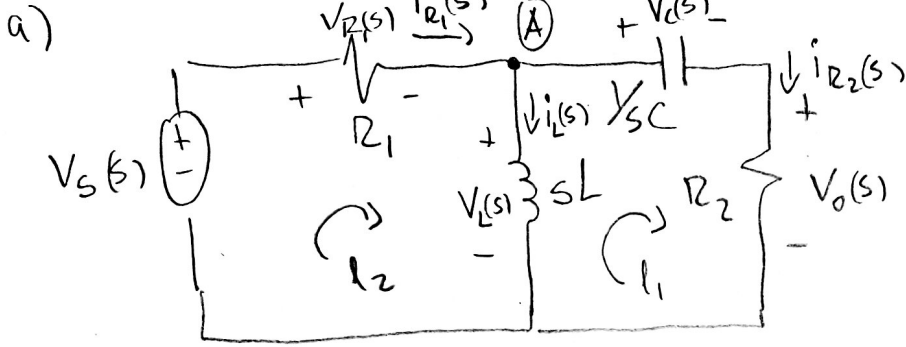
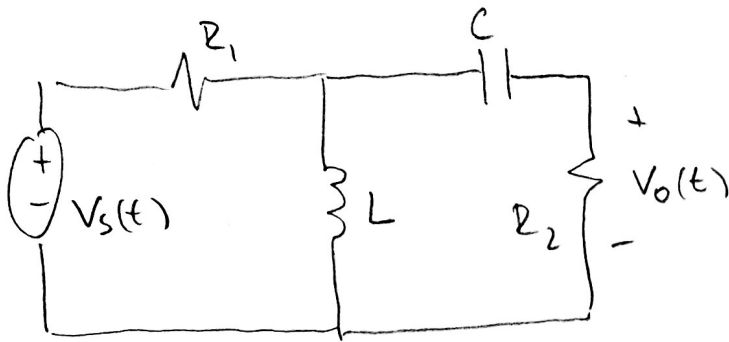
$$\left(10^3 + \frac{10^6}{R_2} + \frac{1}{10^{-3}}\right) = 3000$$

$$\Rightarrow R_2 C_2 = 10^{-3}$$

$$R_2 = (3000 - 2 \cdot 10^3)^{-1} 10^6 = 1 \text{ k}\Omega$$

$$C_2 = 1 \mu\text{F}$$

10.71)



b) We use the unit output method to find $\frac{V_o(s)}{V_s(s)}$:

We assume $V_o(s) = 1 \text{ V-s}$, then

$$i_{R_2}(s) = \frac{V_o(s)}{R_2} = \frac{1}{R_2}$$

$$V_C(s) = \frac{i_{R_2}(s)}{sC} = \frac{1}{R_2Cs}$$

KVL at ℓ_1 : $-V_L(s) + V_C(s) + V_o(s) = 0$

$$V_L(s) = \frac{1}{R_2Cs} + 1 = \frac{1 + R_2Cs}{R_2Cs}$$

$$i_L(s) = \frac{V_L(s)}{sL} = \frac{1 + R_2Cs}{sL(R_2Cs)} = \frac{1 + R_2Cs}{LR_2Cs^2}$$

KCL at node (A):

$$i_{R_1}(s) = i_L(s) + i_{R_2}(s) = \frac{1 + R_2Cs}{R_2LCS^2} + \frac{1}{R_2}$$

$$= \frac{R_2 + R_2^2Cs + R_2LCS^2}{R_2^2LCS^2} = \frac{1 + R_2Cs + LCS^2}{R_2LCS^2}$$

10.71) cont:

$$V_{R_1}(s) = R_1 i_{R_1}(s) = \frac{R_1 (LCs^2 + R_2Cs + 1)}{R_2LCs^2}$$

KVL at l_2 : $-V_s(s) + V_{R_1}(s) + V_L(s) = 0$

$$V_s(s) = \frac{R_1LCs^2 + R_1R_2Cs + R_1}{R_2LCs^2} + \frac{1 + R_2Cs}{R_2Cs}$$

$$= \frac{R_1LCs^2 + R_1R_2Cs + R_1 + Ls + LR_2Cs^2}{R_2LCs^2}$$

Therefore,

$$\frac{V_o(s)}{V_s(s)} = \frac{1}{V_s(s)} = \frac{R_2LCs^2}{(R_1 + R_2)LCs^2 + (R_1R_2C + L)s + R_1}$$

c) We have
$$\frac{V_o(s)}{V_s(s)} = \frac{\frac{R_2}{R_1 + R_2} s^2}{s^2 + \frac{R_1R_2C + L}{(R_1 + R_2)LC} s + \frac{R_1}{(R_1 + R_2)LC}}$$

$$s^2 + 2\zeta\omega_0 s + \omega_0^2$$

with $\zeta = 0.707$ and $\omega_0 = 707 \text{ rad/s}$, we have

$$2\zeta\omega_0 = 2(0.707)(707) = 999.698 = \frac{R_1R_2C + L}{(R_1 + R_2)LC} = \frac{500^2C + L}{1000LC} \quad (*)$$

$$\omega_0^2 = 707^2 = 499849 = \frac{R_1}{(R_1 + R_2)LC} = \frac{500}{1000LC} = \frac{1}{2LC} \quad (**)$$

From (*), we have $999.698 \cdot 10^3 LC = 250 \cdot 10^3 C + L$

and from (**)

$$999698 LC = 1 \Rightarrow C = \frac{1}{999698L}$$

From (***)

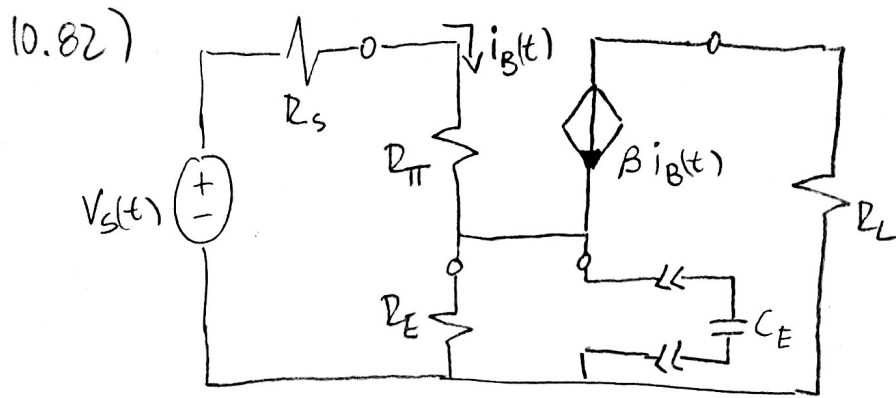
$$1 = \frac{250 \cdot 10^3}{999698L} + L \Rightarrow L^2 - L + \frac{250 \cdot 10^3}{999698} = 0$$

10.71) cont.

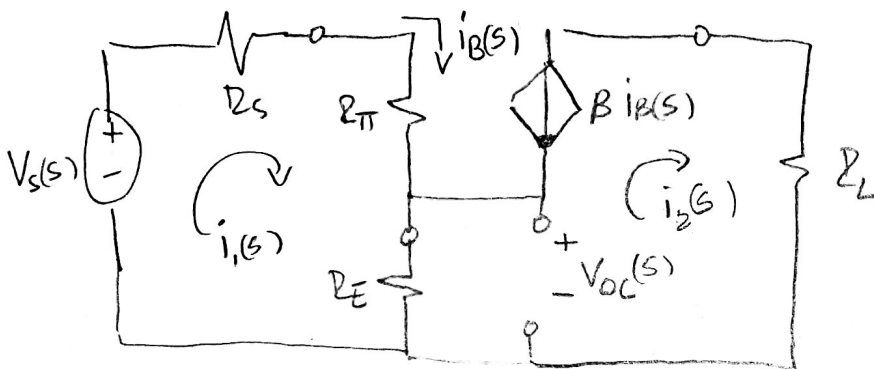
$$L^2 - L + 0.25 = 0$$

$$L = 0.5 \text{ H}$$

$$C = \frac{1}{999698(\frac{1}{2})} = 2 \mu\text{F}$$



To find the Thévenin equivalent, first we find $V_{oc}(s)$ as shown in the next plot:



We proceed by mesh-current analysis:

$$-V_s(s) + (R_s + R_\pi + R_E) i_1(s) - R_E i_2(s) = 0 \quad (1)$$

$$i_2(s) = -\beta i_B(s) = -\beta i_1(s) \quad (2)$$

Where we used the fact: $i_B(s) = i_1(s)$. Next we replace (2) in (1):

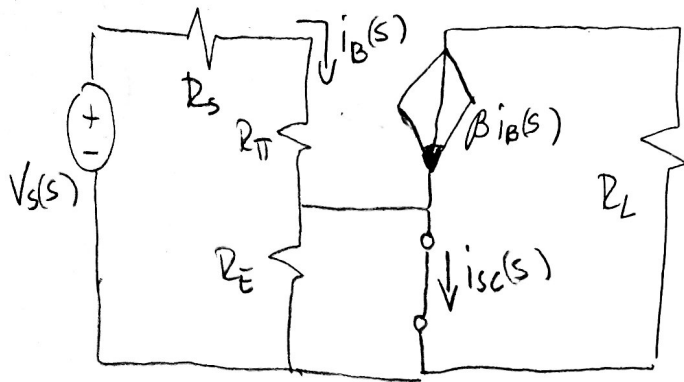
$$-V_s(s) + (R_s + R_\pi + R_E + \beta R_E) i_1(s) = 0 \Rightarrow i_1(s) = \frac{V_s(s)}{R_s + R_\pi + R_E + \beta R_E}$$

Notice that $V_{oc}(s)$ is the voltage in the resistor R_E , then

$$\begin{aligned} V_{oc}(s) &= R_E (i_1(s) - i_2(s)) = R_E (i_1(s) + \beta i_1(s)) \\ &= \frac{(1 + \beta) R_E V_s(s)}{R_s + R_\pi + R_E + \beta R_E} \end{aligned}$$

10.82) cont:

Next, we find the short-circuit current:



$$i_{sc}(s) = i_B(s) + \beta i_B(s) = (1 + \beta) i_B(s) \quad (3)$$

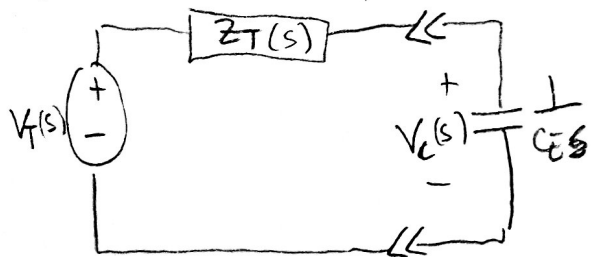
$$i_B(s) = \frac{V_s(s)}{R_s + R_{\pi}} \quad (4)$$

Replacing (4) in (3): $i_{sc}(s) = \frac{(1 + \beta) V_s(s)}{R_s + R_{\pi}}$

It follows that: $Z_T = \frac{V_{oc}(s)}{i_{sc}(s)} = \left(\frac{(1 + \beta) R_E V_s(s)}{R_s + R_{\pi} + (1 + \beta) R_E} \right) \left(\frac{R_s + R_{\pi}}{(1 + \beta) V_s(s)} \right)$

$$= \frac{R_E (R_s + R_{\pi})}{R_s + R_{\pi} + (1 + \beta) R_E}$$

The Thevenin equivalent is



Using the values: $R_s = 10\text{K}\Omega$, $R_{\pi} = 2\text{K}\Omega$, $R_E = 3.3\text{K}\Omega$, $R_L = 1\text{K}\Omega$, and $\beta = 70$, we have:

$$V_T(s) = V_{oc}(s) = \frac{(1 + 70)(3.3\text{K}) V_s(s)}{10\text{K} + 2\text{K} + (1 + 70)3.3\text{K}} = 0.95 V_s(s)$$

$$Z_T(s) = \frac{3.3\text{K}(10\text{K} + 2\text{K})}{10\text{K} + 2\text{K} + (1 + 70)3.3\text{K}} = 160.78\Omega$$

Next, we find the pole associated with the capacitor. For that we use voltage division:

$$V_C(s) = \frac{\frac{1}{C_E s}}{Z_T(s) + \frac{1}{C_E s}} V_T(s) = \frac{V_T(s)}{Z_T(s) C_E s + 1} = \frac{V_T(s)}{160.78 C_E s + 1} \quad (5)$$

10.82) cont

We are given $s = -300$, and from (5) we have

that $160.78 C_E s + 1 = 0 \Rightarrow s = \frac{-1}{160.78 C_E} = -300$

which gives $C_E = \frac{1}{(160.78)(300)} = 2.07 \times 10^{-5} \text{ F}$