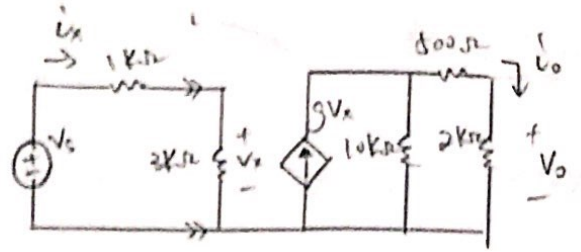


HW5 Solution

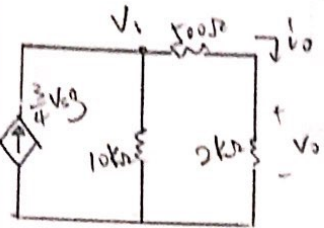
4.3.

$$V_x = i_x \cdot 3 \text{ k}\Omega = \frac{V_s}{4 \text{ k}\Omega} \cdot 3 \text{ k}\Omega$$

$$= \frac{3}{4} V_s$$



then, the right half circuit:



at V_1 :

$$V_1 = \frac{3}{4} V_s \cdot (10 \text{ k}\Omega \parallel (2 \text{ k}\Omega + 500 \Omega))$$

$$= \frac{6000}{4} V_s$$

then, $V_o = \frac{2000}{2000 + 500} \cdot V_1 = \frac{4}{5} \cdot \frac{6000}{4} V_s \Rightarrow \boxed{\frac{V_o}{V_s} = 3.6}$

$i_o = \frac{V_o}{2 \text{ k}\Omega}$ & $i_x = \frac{V_s}{4 \text{ k}\Omega} \Rightarrow \frac{i_o}{i_x} = \frac{V_o / 2 \text{ k}\Omega}{V_s / 4 \text{ k}\Omega} = 2 \cdot \frac{V_o}{V_s} = \boxed{7.2}$

$P_i = V_s i_x = \frac{V_s^2}{4 \text{ k}\Omega} = \frac{100 \text{ V}^2}{4000 \Omega} = \boxed{0.025 \text{ W}}$

$P_o = i_o^2 \cdot 2 \text{ k}\Omega = (7.2 \cdot i_x)^2 \cdot (2 \text{ k}\Omega) = (7.2 \cdot \frac{V_s}{4 \text{ k}\Omega})^2 \cdot (2 \text{ k}\Omega) = \boxed{0.648 \text{ W}}$

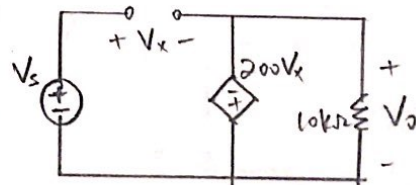
4.6

$V_x = V_s - V_o$

$200 V_x + V_o = 0$

$\Rightarrow 200 V_s - 200 V_o + V_o = 0$

$\Rightarrow \boxed{\frac{V_o}{V_s} = \frac{200}{199}}$



4.30

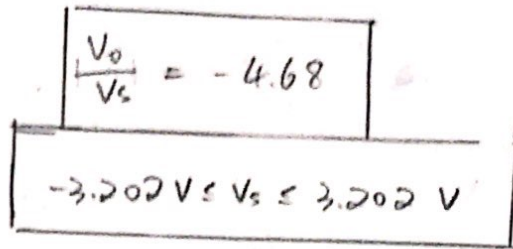
$$\frac{V_A - V_S}{22 \text{ k}\Omega} + \frac{V_A - 0}{47 \text{ k}\Omega} + \frac{V_A - 0}{33 \text{ k}\Omega} = 0 \Rightarrow V_A = 0.468 V_S$$

Inverting amplifier:

$$V_O = - \frac{330 \text{ k}\Omega}{33 \text{ k}\Omega} V_1 = -10 V_1$$

$$\Rightarrow -15 \text{ V} \leq -4.68 V_S \leq 15 \text{ V}$$

\Rightarrow



$$\frac{V_O}{V_S} = -4.68$$

$$-3.202 \text{ V} \leq V_S \leq 3.202 \text{ V}$$

4.38

Without doing any calculation, we see that $V_{CC} = \pm 15 \text{ V}$

so that $V_O \leq 15 \text{ V}$, and we could never reach a gain of 10.

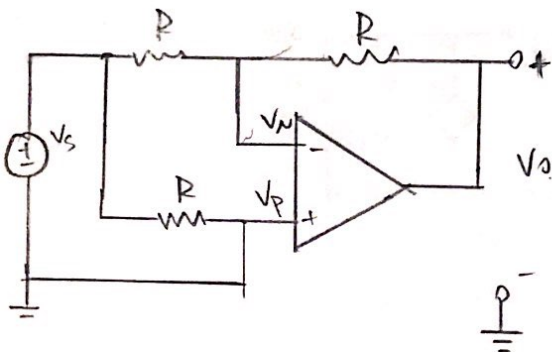
since $10 \cdot 2 \text{ V} = 20 \text{ V} > 15 \text{ V}$ (Assuming op-amp w/ gain = $\frac{R_f + 10 \text{ k}\Omega}{10 \text{ k}\Omega}$)

Case 1: gain = $\frac{0 + 10 \text{ k}\Omega}{10 \text{ k}\Omega} = 1$. Case 2: gain = $\frac{40 \text{ k}\Omega + 10 \text{ k}\Omega}{10 \text{ k}\Omega} = 5$. Case 3: gain = $\frac{90 \text{ k}\Omega + 10 \text{ k}\Omega}{10 \text{ k}\Omega} = 10$

A simple fix is to lower input voltage or increase V_{CC} of amplifier

4.48

When switch is closed,



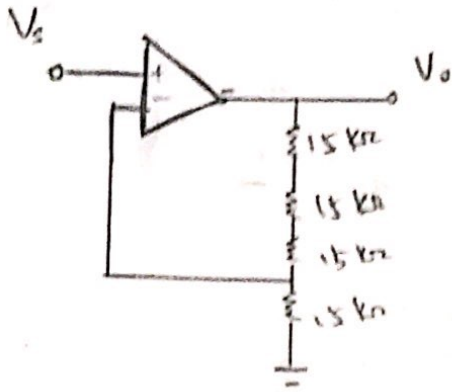
$$V_P = V_N = 0$$

Inverting amplifier:

$$V_O = - \frac{R}{R} V_S = -V_S$$

\therefore claim not correct

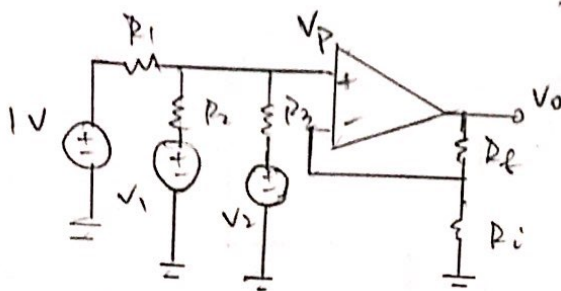
4.63



Note:

multiple answers are correct

4.68.



→ zero out V_1, V_2 .

$$\frac{V_{p1}}{R_1} + \frac{V_{p1}}{R_2} + \frac{V_{p1}}{R_3} = 0$$

$$\left(\frac{1}{R_1} + \frac{1}{R_2} + \frac{1}{R_3}\right) V_{p1} = -\frac{1}{R_1}$$

$$\Rightarrow \frac{\frac{1}{R_1}}{\frac{1}{R_1} + \frac{1}{R_2} + \frac{1}{R_3}} = \frac{1}{3} \text{ by letting } R_1 = R_2 = R_3 = 1 \text{ k}\Omega$$

$$V_{o1} = \frac{R_f + R_i}{R_i} \cdot V_{p1} = \left(\frac{R_f + R_i}{R_i}\right) \left(\frac{1}{3}\right) (1V) \rightarrow \text{let } R_f = 17 \text{ k}\Omega, R_i = 1 \text{ k}\Omega$$

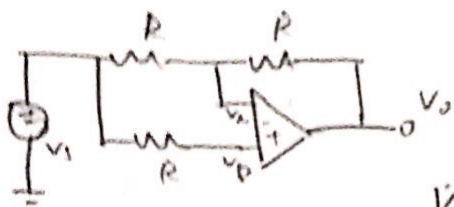
Similarly, zero out V_1, V_2 , the relationship $V_{o2} = 6 V_1$ still holds and for $V_{o3} = 6 V_2$.

$$\therefore R_1 = R_2 = R_3 = R_i = 1 \text{ k}\Omega, R_f = 17 \text{ k}\Omega$$

Note:

multiple solutions possible.

when switch is open



$$V_p = V_n \text{ and } i_p = i_n = 0$$

$$\Rightarrow V_p = V_s = V_n$$

KCL at V_n :

$$\frac{V_n - V_s}{R} + \frac{V_n - V_0}{R} = 0 \Rightarrow V_0 = V_s$$

\therefore claim is not correct

4.50

a) KCL at V_1 :

$$\frac{V_1 - V_2}{R_1} + \frac{V_1 - 0}{R_2} + \frac{V_1 - V_0}{R_3} = 0$$

KCL at V_2

$$\frac{V_2 - V_s}{R} + \frac{V_2 - V_1}{R_1} = 0$$

Also, ideal op-amp $\Rightarrow V_2 = 0 \Rightarrow V_1 = -\frac{R_1}{R} \cdot V_s$

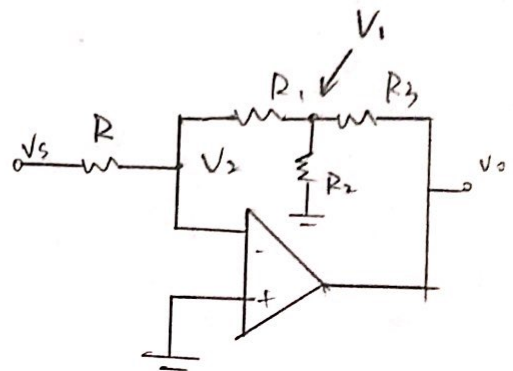
$$\text{then } V_0 = V_s \cdot \frac{-R_1 R_2 - R_1 R_3 - R_2 R_3}{R \cdot R_2} = V_s \cdot K$$

$$\Rightarrow K = -\frac{R_1 R_2 + R_1 R_3 + R_2 R_3}{R \cdot R_2}$$

b) let $R = 1 \text{ k}\Omega$, $R_1 = 1 \text{ k}\Omega$, $R_2 = 1 \text{ k}\Omega$, $R_3 = 4.5 \text{ k}\Omega$

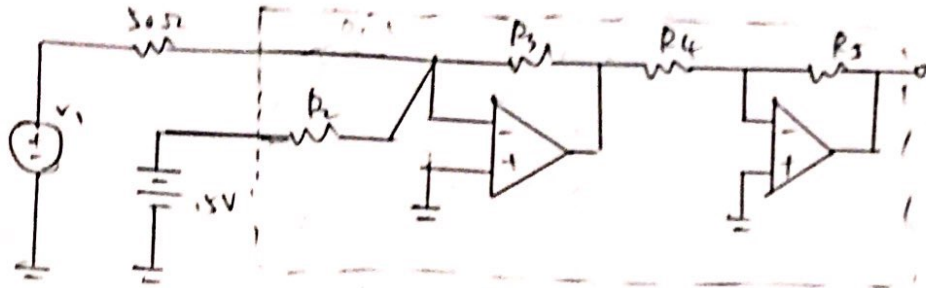
$$\Rightarrow K = -10$$

(multiple solution are possible)



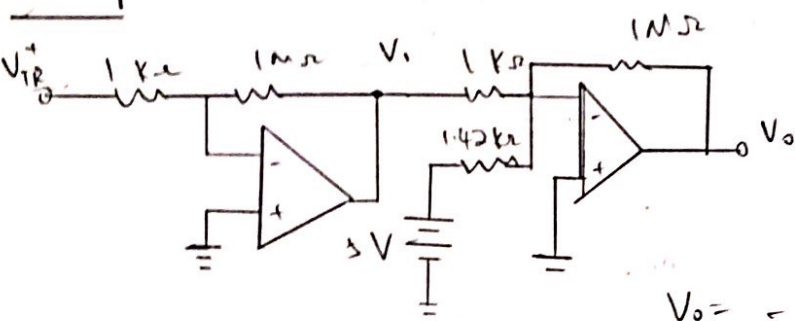
4.73

Design an op-amp circuits with gain of 150 for V_i and 0.1 for 15V source (multiple answer acceptable, using 2 op-amp is easier)



let $R_4 = R_5 = 1k\Omega$
and $R_3 = 7.5k\Omega$
 $R_2 = 75k\Omega$

4.89



$$V_i = -\frac{1M\Omega}{1k\Omega} V_{TR}^+$$

$$= -1000 V_{TR}$$

then, inverting summer

$$V_o = -\frac{10^6}{10^3} \cdot V_{o1} + \left(-\frac{10^6}{1.42 \times 10^3} \cdot 3.5V\right)$$

$$= 10^6 \cdot V_{TR} - 3.5V$$

∴ produces the correct output → could busy

When connecting to transducer, there would be an effect due to the output resistance on the overall gain. Thus, a buffer could be used (for free) to solve this.

